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MULTIMISSION AIRCRAFT DESIGN STUDY: ELECTROMAGNETIC COMPATIBILTIY

THESIS

Jenna M. Davis, Captain, USAF AFIT/GAI/ENY/03-01

DEPARTMENT OF THE AIR FORCE AIR UNIVERSITY

AIR FORCE INSTITUTE OF TECHNOLOGY

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AFIT/GAI/ENY/03-01

MULTIMISSION AIRCRAFT DESIGN STUDY: ELECTROMAGNETIC COMPATIBILITY

THESIS

Presented to the Faculty

Department of Aeronautics and Astronautics

Graduate School of Engineering and Management

Air Force Institute of Technology

Air University

Air Education and Training Command

In Partial Fulfillment of the Requirements for the

Degree of Master of Science in Systems Engineering

Jenna M. Davis, BS

Captain, USAF

March 2003

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Abstract

The multi-mission aircraft (MMA) technical feasibility study looked at the replacement of the aging fleet of C-135 and C-130 theater based command & control (C2) and intelligence, surveillance and reconnaissance (ISR) fleet. It is proposed that the MMA be out-fitted to combine some or all the functions of existing AWACS, JSTARS, RIVET JOINT, COMPASS CALL, and ABCCC platforms. It would also have links to other manned or unmanned ISR aircraft, as well as satellites.

The objective of the proposed design study is to examine the technical risks involved in combining multiple functions onto one aircraft that currently reside on separate aircraft. This thesis specifically focused on the risks that are due to electromagnetic interference between transmitters and interference between active and passive sensors.

Two architectures were examined: one tail number (OTN) and different tail number (DTN). The OTN architecture was found to be incompatible due to interference between the air moving target indicator transmit and high band receive functions, whereas, the DTN was found to be compatible for all variant architectures.

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I. Introduction

Background

Tasking

Major General Glen D. Shaffer, Director for Intelligence, Surveillance and Reconnaissance (ISR), DCS, Air and Space Operations, United States Air Force (USAF) has requested a technical feasibility study for a multi-mission aircraft (MMA). According to Major General Shaffer, the MMA concept has been proposed as a replacement for the aging fleet of C-135 and C-130 theater-based command and control (C2) and intelligence, surveillance and reconnaissance (ISR) fleet. It is proposed that the MMA be out-fitted to combine some or all the functions of the existing AWACS, JSTARS, RIVET JOINT, COMPASS CALL, and ABCCC platforms. It would also have links to other manned or unmanned ISR aircraft, as well as satellites.

<u>Objective</u>

In performing a MMA feasibility study, the primary goals are to replace the current aging fleet with a single platform. Reduced life cycle costs, increased system value through measure of mission utility and mission integration and compatibility, and minimal risk are the primary objectives considered. The overall need is to ensure that every mission currently being served by this fleet will not only continue but also enhance a theater's ability to perform time critical targeting (TCT).

To consolidate the platforms, we must first understand current mission requirements. The AWACS is in charge of air moving target indication (AMTI), weapons C2, air battle management (ABM) and identification of friend or foe (IFF). The

JSTARS provides long-range ground moving target indicator (GMTI) surveillance, synthetic aperture radar (SAR) surveillance and wide area search (WAS), ground C2, and ground battle management (GBM). RIVET JOINT provides ISR information and electronic warfare support to theater commanders (electronic battlefield management). COMPASS CALL provides primarily air C2 and communications countermeasures (C3CM) but can provide jamming support to ground forces. The ABCCC is the overall tactical command and control.

Preliminary Analysis

Preliminary Group Design

The investigation of the multimission aircraft began with a preliminary group of twelve students comprised of logistics and maintenance operations, air and space operators, and acquisition, science and engineering backgrounds. The preliminary group brainstormed and researched the current platforms to develop two baseline hierarchies and value system designs (VSD) using Hall's Seven Steps.¹ In addition, a concept map (Appendix 1.2) was constructed to show relationships between key players, systems and operational considerations.

Based on an interactions matrix similar to the matrix in Appendix 1.3, an interface and flow model (Appendix 1.4) were created using techniques defined by Hatley, Hruschka and Pirbhai (HHP)². The HHP techniques help to stimulate system specifications to iteratively generate a set of system requirements and architecture models. The interface model depicts key requirements and interactions within the MMA

¹ Hall's Seven Steps will be discussed in detail in Chapter 3. Methodology.

² Hatley, Hruschka and Pirbhai (HHP) methodology will be discussed in Chapter 3: Methodology.

design. The flow model was then used to build and track the architectures and some of their variants. The process interface was the centerpiece or driving force behind the iterations. As each architecture was developed, the system requirements were enhanced and fed back into the interface. As the process continued, several architecture variations developed and are noted as sub-bullets in the systems architecture model.

MMA Thesis Team

The MMA thesis team consisted of a group of three students including Lieutenant (LT) Nevin Coskuner, Turkish Air Force (TUAF), LT Ahmet Kahraman, TUAF, and myself. The MMA thesis team reinvestigated, compiled and developed a new and complete baseline including a systems definition consisting of key players, stakeholders, needs, alterables and constraints. An interaction matrix was developed based on these system definitions to visually show cross-interactions.

The interaction matrix found in Appendix 1.3 was a key element to building the system synthesis architecture as it identified where special or in-depth research was needed to accomplish an understanding of the system design to the fullest extent. To logically assign levels of interaction, the designated strengths; high, medium, and low, were assigned numerical values. Each element value was totaled based on its interaction among the other elements. For each group (objective, alterable, constraint and need), the elements were arranged in order, based on this total, and natural group interaction levels were established. The cross-interactions have been summarized and categorized by level in Table 1. The analysis of the interaction-matrix determined the system variables that drove the design to the most or at the "highest level." Other interaction levels were addressed as needed.

	HIGH INTERACTION	MEDIUM INTERACTION	LOW INTERACTION
OBJECTIVES	Max Mission Effectiveness Mission Integration & Compatibility	Minimize Risk	Minimize LCC
	Air C2 Ground C2	ISR Processing & Exploration Air BM	Longterm Compatibility
NEEDS	ISR Collect. & Recog. Mission Dissemination & Transmission	Ground BM C3 CM Joint Service Interoperability	All-Weather Capability (24/7)
ALTERABLES	System Architecture	Mission Requirements	Future Politics/Players/ CONOPs
CONSTRAINTS	Operations Environment Tecnnology Availibilty	Air Frame Limits Funding	Safety
CONSTRAINTS	Development Time System Compatibility	Classification of System Logistics Supportability	Gov't regulations & Policies

 Table 1: Objectives, Needs, Alterables and Constraints Summary by Level of Cross-Interaction.

Areas of Investigation

The MMA thesis team determined three key areas³ for further investigation from the cross-interactions of the system definition constraints. These areas consisted of: 1) payload limitations based on airframe limits, 2) the operations environment, and 3) system compatibility. The One Tail Number (OTN) and Different Tail Number (DTN) architectures will be examined under these emphasis areas. The DTN architecture will consist of four alternative architectures.

Aircraft Design as it Pertains to Payload Limitations

In order to give specific answers for a MMA design and its compatibility, we should be aware of what is going to be integrated into the MMA architecture. Basically, we can say that those should be the sensors for the joint missions, the crew, and all of the software and the hardware for the missions. By investigating aircraft payload integration, we will be able to make decisions based on key factors such as weight, volume, range, and some other related limits of the aircraft. To accomplish this, we need an

³ Chapter 4. Process Tailoring and Results will include a more detailed discussion of how the levels of interactions were determined.

understanding of the sensors and antennas mass and volume characteristics. We will then be able to make decisions about the compatibility of the two architectures and their variants. LT Kahraman, TUAF, is accomplishing this research. (Kahraman)

Operations Environment Design Parameters

By assigning all of the C3CMISR missions capabilities under one aircraft, the requirements may prove to be too diverse and cover too large of a defined mission area for a single MMA aircraft. Thereby reducing the purported advantage of consolidating the capabilities. As the Area of Interest (AOI) and/or the number of multiple taskings grow, the mission effectiveness may decrease along with overall performance. It is for these reasons that the operations environment is believed to be a key decision area, as it will affect the concept of operations, logistics and C2 and ISR areas of coverage. This portion of the research will develop a hypothetical conflict area with a defined set of constraints by which the OTN and DTN architectures will be evaluated. Lt Coskuner, TUAF, is accomplishing this research. (Coskuner)

Payload Integration as it Pertains to Electromagnetics

By the Department of Defense, Joint Pub 1-02, the electromagnetic environment effects (E3) is defined as

The impact of the electromagnetic environment upon the operational capability of military forces, equipment, systems, and platforms. It encompasses all electromagnetic disciplines, including electromagnetic compatibility/electromagnetic interference (EMC/I); electromagnetic vulnerability; electromagnetic pulse; electronic protection, electromagnetic radiation hazards to personnel, ordnance, and volatile materials; and natural phenomena effects of lightning p [precipitation]-static

As the specific mission aircraft are integrated, the discipline of EMC/I will be a key concern. This key element must be understood completely before development

begins or else there will be a higher potential for unintentional interference throughout the system. The emissions, attenuation, power influences, shielding influences, antenna placement, radiation and characteristics of the C3 and ISR equipment were a few specific areas that were investigated.

<u>Scope</u>

The intent of this thesis is to develop and apply a first order model focused on the primary decision variables and parameters that allow an evaluation of the impacts of electromagnetics on MMA configuration. A preliminary EMC analysis developed by Don White Consultant will be used to determine potential antenna-to-antenna interference among the major systems.

EMC/I impacts on the MMA value system design will be discussed and a summary of the system design including the results from Lt Kahraman's thesis. The results of this work are intended to give additional insight into the ongoing Multimission Command and Control Architecture (MC2A) and to hopefully provide ideas or thoughts not considered before.

Assumptions/Limitations

All of the aforementioned systems are US classified systems and will only be referred to as the job that each system performs. Each aircraft platform was given a generic system performance description based on open literature information and notional data. These system parameters can be found in Appendix 3.3. The performance data will then be generated based on typical values for each type of sensor. An example is the sensor frequency where the mean of the community standard range for each asset will be

used to generate a recommended architecture. It will be left to the end user to evaluate the decision model at the properly assigned spectrum frequency.

Based on potential interference severity levels defined by J.L. Wilson and W.B. Jolly⁴, only antenna-antenna radiated coupling will be evaluated for the consideration of EMC.

For a complete EMC analysis, each transmitter-receiver pair would need to be analyzed. In this study, only the air moving target indicator (AMTI), ground moving target indicator (GMTI), low frequency (LF) receiver, high frequency (HF) receiver, and super high frequency (SHF) receiver will be analyzed. The communications links are assumed to work with all architecture combinations. In reality this assumption is more than likely not feasible but the inclusion of the communications architecture would be overwhelming with the consideration that each combination would need to be analyzed. This detailed analysis will therefore be left to a person specializing in EMC/I. With this said, ABCCC is strictly considered a communications node and will not be analyzed along with the mock systems of AWACS, JSTARS and Rivet Joint.

Terminology

Throughout the paper multimission aircraft (MMA) and multimission command and control architecture (MC2A) will refer to the architecture under investigation and will be used interchangeably. Command, control, communication, countermeasures, and intelligence, surveillance and reconnaissance (C3CMISR) will be referred to as the mission requirements to be performed by the MC2A.

⁴ Levels of EMI severity involved with modifying various command, control, communications and intelligence systems (C3I) as described by Wilson and Jolly are shown in Table 4.

Preview

Chapter Two of this thesis describes the current systems engineering approaches, EMC background, and discusses ongoing multimission aircraft development activities in the United States Air Force (USAF). Chapter Three presents the methodology employed in the study. Chapter Four describes and analyzes the resulting model. Chapter Five provides conclusions and recommendations based on the model

II. Literature Review

Systems Engineering Process

The what, how, and methods of facilitation of systems engineering have evolved through time with the creation of processes, modeling techniques, and tool development, respectively. The process defines what is to be done by establishing a logical sequence of tasks. In the 1970's, the waterfall process was the primary construction element of systems processes. Designs like A. D. Hall's three-dimensional morphological box, Space Mission Analysis and Design (SMAD), and System Engineering Process by INCOSE are based on this pseudo-iterative, one directional flow approach. Each of these processes places emphasis on different areas of the development process. The Hall's morphological box process focuses on project planning, value system design and alternative design and analysis. While the SMAD process deals with concept exploration for detailed physical development requiring a considerable amount of upfront planning. Lastly, the INCOSE process primarily deals with the development, production test, deployment, training, support and disposition. The INCOSE process is slightly different from the previous two processes in that it looks at concurrently developing the design layers of the system and looks at the external and enterprise environmental factors.

In the 1980's, the community began to refine the process via multiple iterations referred to as the spiral development. The 1990's made way for two-way interactions. No longer was the thought of a project a direct flow from the beginning to the end product. Instead one could start the process from the bottom, middle or top and enhance the detail as appropriate. This was the beginning of the processes and methods based on structured analysis such as Hatley/Pirbhai methods.

The start of the 2000's brought even more enhancements to these processes with design of the architecture being developed alongside the requirements. The Process for System Architecture and Requirements Engineering (PSARE) also known as the Hatley, Hruschka and Pirbhai (HHP) methods and the integrated definition for function modeling (IDEFO) as described by Dennis Buede are two examples of this era. (Buede, Hately)

Several other processes have been developed during each of the time periods discussed above. The Hall's morphological box, SMAD, and PSARE will be discussed in detail, as they will be used in the methodology of this study.

Although these processes have evolved over time, each process is still viable and implemented and used today. No one process could adequately describe all possible situations or studies. The choice must be based on the end product or the type of study to be performed. The final process could even be a combination of a number of different methods based on the final goal of the study. The strengths of a few processes could be combined to create a tailored process.

There are basically two types of studies: feasibility studies and studies with a product implemented. A feasibility study focuses on needs, alterables and constraints to develop alternative architectures and recommendations for implementation based projects. A detailed value system design is established. However, an overall lack of emphasis on system requirements exists.

Contrary to the feasibility study, the studies with a product to be implemented focus on requirements development, cost analysis, performance and risk. A value system design is not needed for architecture evaluation because only one defined architecture exists. In this case, the value system becomes the constraint for the architecture.

In the end each study must use a systems engineering process which is logical, repeatable and defendable for designing and or selecting a system to answer the study in question (SENG 520 Notes).

Hall's Morphological Box

Hall's morphological box's vertices are comprised of the logic in which the process is to be carried out, the phases of time that occur throughout the development, and the knowledge base of which information is derived from specialized disciplines.

Both the phase and logic component are comprised of seven elements. The one directional flow begins with the first step of the phase structure and works right through the logic structure. Once all of the logic steps have been accomplished for the current phase, the phase advances and the logic steps are reaccomplished. Iterations should be continuously performed within each phase and the process should be advanced to the next phase once all logic steps have been thoroughly evaluated (Sage: 3-4; Hill: 610-611). Halls's activity matrix in Table 2 outlines the relationships between the logical steps and phases elements of the system's engineering process.

Steps of the Fine Structure LOGIC Phases of the coarse structure TIME	Problem Definition	Value System Design	System Synthes is	Systems Analysis	Rank (optimize) Alternatives	Decision Making	Planning for Action
Program Planning							
Project Planning							
System Development							
Production							
Distribution							
Operations							
Retirement							

 Table 2. Hall's activity matrix. (Sage: 5; Hill: 611)

The third dimension of the morphological box, knowledge, is a very important aspect of the process. This knowledge dimension is especially important for the problem definition that should be accomplished as a group activity. This group should be comprised of the stakeholders, the functional engineers, and policy, government, and management specialists. At the beginning of the study, the overall system manager should ensure that all disciplines required for the project are represented. This helps to prevent individual biases based on personal perception to not be incorporated into the system. The assortment of specialties included in the group will also allow for a more complete or total picture of the situation (Lendaris: 604).

<u>SMAD</u>

James Wertz and Wiley Larson address requirements development based on values and objective structuring in their text entitled *Space Mission Analysis and Design*. SMAD is a process very similar in order and content to the Hall's morphological box.

The main exception being that SMAD focuses on the steps of a feasibility study by focusing on the performance objectives (needs), constraints, and concept exploration. These items are generally investigated during the first phase of Hall's process. In addition, the SMAD process primarily focuses on one architecture and performs feasibility analysis at decision nodes of the design development. In doing this, the SMAD process doesn't have a need to concentrate on a value system design. (SENG 520 notes)

SMAD's equivalent to the Hall's knowledge axes specifically includes the inputs of the operator, user and developer to ensure a more realistic and affordable end product. Table 3 shows the space-focused process that has been continuously iterated on over the past 40 years.

Define	Step 1.	Define broad objectives and constraints.		
Objectives	Step 2.	Estimate quantitative mission needs and constraints.		
Characterize	Step 3.	Define alternative mission concepts.		
the Mission	Step 4.	Define alternative mission architectures.		
	Step 5.	Identify system drivers for each.		
	Step 6.	Characterize mission concepts and architectures.		
Evaluate the	Step 7.	Identify critical requirements.		
Mission	Step 8.	Evaluate mission utility.		
	Step 9.	Define mission concept (baseline).		
Define	Step 10.	Define system requirements.		
Requirements	Step 11.	Allocate system requirements to elements.		

Table 3. The Space Mission Design and Analysis Process. (Wertz : 2).

The SMAD process is an iterative approach. In general one would work down

from step 1 to 6. At steps 7-11, one could choose to continue on or flow back to any of

the first 6 steps.

<u>PSARE</u>

The PSARE process consists of three major blocks forming a closed loop: 1)

external stakeholders, 2) system in service, and 3) system development project blocks.

These blocks are composed of a network of elements each having equal status and therefore, are of no particular sequence. This is generally referred to as a concurrent development process. The most significant difference between the PSARE process and the previously mentioned processes is that the PSARE addressed both the requirements development along with the architecture development. In analyzing the requirement and architecture elements together, the essential problem (the what) and constraints imposed on the system (the how to solve) are concurrently developed. This allows for extremely complex system construction to be manageable, and upgrades and/or the reuse of current technologies to be easily integrated.

The PSARE process is outlined in Figure 1. Within the deliverable system development, the system layer addresses the overall structure of the system model. The top system element level further decomposes the individual elements of the system layer. The exact system technology to be used on a specific function configuration is established in the system technologies configuration layer. The system technologies configuration maps the structure of the architecture to the real physical component. The last layer, implementation, mostly consists of detailed design. Each of these layers produce specifications which are fed into a sub-layer and/or into the integration and test development. The integration and test development phase helps to identify constraints on the system. The issues are fed back into the deliverable system development or the completed product is pushed to the system in service to field test/operate.



Figure 1. The Total System Life Cycle (Hatley: 182)

Electromagnetic Compatibility and Interference

Electromagnetic Compatibility (EMC) is the ability for a collection of independent electrical systems to perform without degradation or malfunction to one another in the system's given electromagnetic environment. Electromagnetic Interference (EMI) is the amount of intentional or unintentional degradation inflicted upon one electrical system by another. In general, there are two types of EMI considered: intrasystem and intersystem. In intrasystem EMI degradation is caused within a system by the system itself. Intersystem EMI is when the conflict is introduced from the surrounding environment in which the system resides. The focus of this research will be on intrasystem compatibility. However, this is not as straightforward as one would think. When discussing the antenna-to-antenna interference characteristics of intrasystem compatibility, the analysis becomes similar to intersystem compatibility. The antennas are actually a part of the internal system, however, their impacts on one-another is via the outside environment.

Conducted interference and radiated interference are two subdivisions within the system EMI that describe the wave-coupling paths. When power is directly transferred via physical connection, the coupling is referred to as conducted. Usually this coupling path is via cabling or wires within a box or system that guide the waves. When a wave is unguided and transferred without physical contact, the path is radiated. The general interference paths are defined by how electromagnetic energy travels from the source to receptor. The radiated paths are:

- Wire-Wire (Cable-Cable): wires (cable) or wire (cable) bundles in close proximity to one another
- Antenna-Antenna: power transmitted from one antenna is received at the port of another antenna exceeds the receptor's susceptibility
- Box-Box: individual black box systems leak power into the vicinity of another system
- All combinations of above: Wire-Antenna, Antenna-Wire, Antenna-Box, Box-Antenna, Wire-Box, and Box-Wire

In a technical report produced by J.L. Wilson and M.B. Jolly, the potential severity of interference inflicted by each of the preceding radiated path combination is (Wilson: 8):

with widding mg various Cor Subsystems (winson: of						
Interference Potential		Equipment as Source of Receptor of EMI on Baseline C3I System				
		Antenna	Cable	Box	Power	
					System	
	Antenna	Slight to	Slight to	Minimal	Minimal	
		Severe	Severe			
Candidate	Cable	Slight to	Slight to	Minimal	Minimal	
Equipment		Moderate	Moderate			
Modification	Box	Minimal	Minimal	Minimal	Slight to	
on C3I System					Moderate	
	Power System	Minimal	Minimal	Slight to	Slight to	
				Moderate	Moderate	

 Table 4. Levels of EMI Severity Involved

 with Modifying Various C3I Subsystems (Wilson: 8)

As can be seen from the table, the radiated path combinations are not expected to impact EMI equally. The dominating path combination is dependent on the system under investigation. "Consequently, in many cases the nine possible radiated coupling paths…reduce to antenna-to-antenna, antenna-to-cable, cable-to-antenna, and cable-to-cable" (Violette: 150). This investigation will cover antenna-to-antenna radiated coupling interference analysis.

Wilson and Jolly include power as an important factor to be paid attention to in addition to the radiated paths. The power supplied may not be able to meet the performance requirements of the multiple systems. Long duration demands may overstrain the power supply causing operational failure during or after the mission (Wilson: 9).

Antenna-Antenna Power Spectral Density

EMI potentially occurs when the power spectral density transmitted from one antenna and received at the port of another antenna exceeds the receptor's susceptibility. Power spectral density is the description of how the average power signal is distributed in frequency due to a one-ohm resistor load (Weiner: 1). An antenna's transmit and receive wave signature potentially creates several types of EMI. A transmit wave contains the fundamental or passband frequency and harmonic emissions. A receive wave generally consists of the fundamental and spurious radio frequencies. Overlap of any of these wave components potentially results in EMI. There are three standard types of transmit-receive EMI: 1) co-channel, 2) adjacent channel (intermodulation, transmitter noise, etc.), and 3) out-of-band (Duff: Vol. 7, 2.2; Wilson: 7, 9). The three transmit-receiver EMI types are graphically depicted on the left side of Figure 2.

In co-channel EMI, the fundamental frequencies directly line up within "plus or minus one-half the narrowest [intermediate frequency] IF bandwidth" (Duff: Vol. 7, 2.2). The adjacent channel is similar to the co-channel except that the fundamental frequencies do not directly line up. Instead, the passband or falloff of the main frequencies may overlap. Adjacent channels can occur over a broad range of frequencies. However, the receiver is generally not sensitive to these outlying frequencies and is only investigated for collocated systems (i.e. same aircraft). The potential co-channel and adjacent EMI types are generally measured as a fundamental interference margin (FIM).

Adjacent channel EMI typical results are intermodulation and broadband transmitter noise. Intermodulation can occur when two or three power spectrum peaks (fundamental, spurious, or harmonic) interact to create a third or fourth peak that lay inband to the fundamental transmitter or receiver power spectrum. The linearity of the surface material can also cause passive intermodulation affects (Weston: 586-587; Wilson: 9). "Third harmonic and third order intermodulation products are the most likely to cause problems (i.e. 3f, 2f1 +/- f2, 2f2 +/-f1) (Weston: 586)."

Lastly, the out-of-band EMI occurs when the: transmitter fundamental overlaps with the receiver spurious, the transmitter harmonic overlaps with the receiver fundamental, and/or the transmitter harmonic overlaps with the receiver spurious. These potential levels of EMI are measured by the transmitter interference margin (TIM), receiver interference margin (RIM), and spurious interference margin (SIM), respectively. Figure 2 graphically shows the EMI measurements on the right side.



Figure 2. Types of Transmitter-Receiver EMI and the Respective Measurements (Duff: Vol.7, 2.3; Violette: 137)

For each depiction in Figure 2, the receiver power density (main frequency input and spurious responses) and the transmit power density (main frequency output and harmonic responses) are represented by the top and bottom signature, respectively. The FIM measurement directly corresponds to co-channel and adjacent channel EMI. Whereas, the RIM, TM and SIM measurements align with the out of band EMI.

Antenna-Antenna Interference Margin

The FIM, RIM, TIM, and SIM are each a special case of the interference margin (IM). The IM is the potential for the transmitted power available at the receiver to exceed the susceptibility threshold. If the IM is positive the likelihood of interference is positive. However, if the IM is negative there is little to no chance of interference. When the two are equal there is a marginal chance that interference may or may not exist.

The energy of the transmitted wave changes as it propagates from one point to another due to the loss of some of the energy into the atmosphere and the accuracy of the pointing direction. Therefore, the relationship must be corrected to show these effects in the following way: (Duff: Vol. 7, 2.8-.10)

$$IM (f,t,d,p) = I/N =$$

$$P_T(f_E) + G_{TR} (f_E,t,d,p) - L (f_E,t,d,p) + G_{RT} (f_E,t,d,p) - P_R (f_E)$$

$$+ CF (B_T, B_R, delta f)$$
(2.1)

where,

I/N is the interference-to-noise ratio

 $P_T(f_E)$ is the power transmitted in dBm at f_E

 G_{TR} (f_E, t, d, p) is the transmitter antenna gain in dB at f_E in the direction of the receiver

 $L(f_E,t,d,p)$ is the propagation loss in dB at f_E between transmitter and receiver

 $G_{RT}(f_E, t, d, p)$ is the receiver antenna gain in dB at f_E in the direction of the transmitter

 $P_R(f_R)$ is the receiver susceptibility threshold in dBm at f_R

CF (B_T , B_R , delta *f*) is the correction factor in *dB* to account for B_T , B_R , and delta *f*

 f_E is the emission frequency

 f_R is the response frequency

t is the time dependency

d is the distance between the transmitter and receiver

p is the polarization of the wave

 B_T is the transmitter bandwidth B_R is the receiver bandwidth delta f is the absolute difference between the transmitter and receiver bandwidths

The way the wave propagates from the transmitter to the receiver determines the propagation loss. The waves can travel directly (for co-site antenna, directivity is modified by a reflection correction), by reflection (which is a function of the conductivity or permittivity of the reflectance surface and the angle of incidence), by surface coupling and by bouncing off the particles in the sky (Weston: 579-580). The pointing direction correction is based on the gain and bandwidths of the transmitting and receiving antennas.

Interference Margin Independent and Dependent Variables

Frequency, time separation, distance, and direction are the independent variables of EMI. The transmit and receive antenna equipment type, age, maintenance condition, and seasonal, environmental and/or atmospheric parameters influence the independent variables. Frequency is the best control for EMI but spurious emissions at other frequencies are hard to control and increase the overall complexity of the EMC problem. Each transmit-receiver pair must be considered in the selectivity and analysis. The antenna rotation, scanning and moving equipment, solar cycles, diurnal effects, seasonal effects and operations influence time separation. The most challenging and/or highest level of expected EMI problems are dealt with in the near-field region. This near-field region is where the determination of minimum distance separation occurs. Direction is the last independent element and is described as the three-dimensional direction and polarization of the electromagnetic wave (Duff: Vol. 7, 2.5-.6).

The variables directly contributing to interference are amplitude, power transmitted, transmission coupling function and the susceptibility threshold. The following relationships summarize Duff's description of the dependent variables (Duff: Vol. 7 2.6-.8):

- Amplitude can be used as a 'weeding-out' process.
- As P_T increases, the potential for interference increases.
- As transmission coupling increases, the potential interference increases.
- As L decreases, the potential interference increases.
- As PR decreases, the potential interference increases.

Antenna-Antenna Modeling Techniques

Modeling of antenna-antenna EMI begins with the electromagnetic characteristic definitions. The receiver characteristics needed to determine potential degradation include the operating frequency range, demodulation process, susceptibility, sensitivity, and IF band input density spectrum. The operating frequency range, type of modulation, modulation bandwidth, and the power density spectrum are required for the transmitter input. The next step is to include any attenuation changes to the transmitter such as any in-line filters. A comparison of the EMI system characteristics is accomplished using a math model. This math model is then used to determine the radio frequency (RF) margins for all transmit-receive combinations. Based on the final IM result, the designer can assess what, if any, corrective measure is needed.

In most of the EMI prediction, the structural coupling path, antenna gain, transmission system mismatch, transmission emission spectrum, and the receive susceptibility response are all sub-model components of the mathematical models. In

addition, operational doctrine and data handling strategies are sometimes modeled. Wilson and Jolly thoroughly discuss the five modeling attributes along with a discussion of the computerized models utilizing these attributes and some of the inaccuracies and difficulties of the computerized models utilizing these attributes. The reader is referred to Wilson and Jolly's paper for further discussion.

This paper will utilize the "Short Form EMI Prediction" tool developed by Don White Consultants in 1972. *The Electromagnetic Compatibility Handbook* by Violette, White, and Violette and *The Handbook Series on Electromagnetic Interference and Compatibility* (Vol. 7) by Duff both discuss this short form model in detail (Violette, Duff).

The form consist of five parts: 1) the FIM, SIM, TM, RIM quick-look, 2) Amplitude, 3) Frequency, 4) Detailed parameter analysis, and 5) Performance Analysis. The form is first used as a preliminary analysis tool to quickly look at and remove low probability EMI cases. After the quick-look, the remaining four analysis levels are considered for further investigation. As each of the remaining level of analysis is performed, 90 percent of the non-interfering situations should be removed (Duff: Vol. 5, 2.17). If applicable, the FIM, TIM, RIM, and SIM cases are tested in each level. Advancement of the antenna-antenna pair into the next level is based on failure to meet the baseline IM requirement.

In the amplitude analysis, the transmission loss is assumed to be minimized and the "emission output and receptor response are aligned in frequency such that the vulnerability device provides minimum rejection to the potential interference signal
(Duff: Vol. 7, 2.17)." Amplitude analysis solves the IM equation with rough propagation loss estimates and doesn't consider the correction factor.

The frequency analysis uses the amplitude results and incorporates the transmitter bandwidth and modulation, bandwidth and selectivity of the receiver, and the frequency separation between the antenna-antenna pair. Each type of transmitter-receiver EMI is compared. In short, this step adds the correction factor into IM equation. "The results of frequency analysis yield surviving cases that have a significant potential for producing interference (Duff: Vol. 7, 2.28)."

Detailed Analysis incorporates the time, distance and direction independent

variables and determines the interference probability distribution. Finally, the

performance analysis measures the signal-to-noise ratio to identify and determine the

extent of damage to the operational performance.

"Short Form EMI Prediction" Assumptions and Limitations

Ten assumptions are defined for the process and suggested values in the form.

The assumptions are (Duff: Vol. 7, 2.38-.39):

1) Frequency limits for transmitter spurious emissions and receiver spurious responses are from 0.1 to 10 times the fundamental frequency. This assumes that there are no significant emissions or responses outside these limits.

2) Maximum TX-RX [transmit-receive] frequency separation for fundamental interference is 0.2 times the receiver fundamental. This assumes fundamental interference is not significant for larger frequency separations.

3) Free-space propagation loss is assumed.

4) Levels for transmitter spurious emissions are 60 dB below fundamental emission.

5) Levels for receiver spurious susceptibility are 80 dB above fundamental susceptibility.

6) An additional 20 dB rejection each is assumed for transmitter and receiver minor emissions and responses.

7) Values for antenna gains in unintentional radiation directions and at unintentional frequencies are 0 dB.

8) Differences in transmitter and receiver bandwidth are assumed to modify the power available in the manner specified in Table 2.1 of Duff [bandwidth corrections in dB].

9) Frequency separation delta f between transmitter emission and receiver response are assumed to reduce the effective power available by an amount given by $40*\log(0.5 [B_T + B_R])/\text{delta f.}$

10) A go, no-go interference margin level of -10dB is used. Thus, potentially interfering situations are eliminated only if the mean signal level is less than -10 dB relative to the receiver susceptibility threshold.

The "Short Form EMI Prediction" tool has many limitations. Some of the biggest

issues with the short form are the geometry assumptions of a flat plane and not a cylindrical surface. Additionally, the short form is not an automated process. This is not a factor for a small number of antenna combinations, however, when the antenna number is large this is a fairly time consuming and inefficient process. Several automated tools that utilize a cylindrical geometry exist. However, these tools require a higher level of detailed input data. For this level of analysis with the fictitious data set, the "Short Form EMI Prediction" tool will be adequate. J. L. Wilson and M. B. Molly created detailed explanations of several automated tools along with the associated modeling attributes and suitability. Although the readers are encouraged to read the detailed discussion themselves, this table has been included in Appendix 3.2 for easy reference.

EM Mitigation Techniques

For the failing transmit-receive pairs, several techniques exist. The type and level of IM determine difficulty and ultimately the incurred costs to perform the correction. The frequency, time, angle, and location are the drivers for control. Frequency management can be used by adjusting transmitter modulation bandwidth, pulse rise and fall time, addition of harmonic filters, and frequency allocation and assignments.

Additionally, the receiver EMI impacts can be controlled by the addition of preselectors, filters and correlators. Time-sharing, radar pulse synchronization and time/range gate controls are examples of time management techniques. Direction management can be implemented by controlling the azimuth and elevation use and assignment, sector banking, space filters and polarization. These techniques are all identified by Duff under intersystem interference and control; however, these can be used for the intrasystem antenna-antenna co-location systems (Duff: Vol. 1, 1.23).

As for intrasystem EMI control, Duff breaks the management into five categories: 1) circuits and components, 2) filtering, 3) shielding, 4) wiring, and 5) grounding. The subcomponents of these categories include arc suppression, power main filters, housing material and thickness, packaging seals and gaskets, cable grouping and grounding, connector shields, and structure and bond grounding, etc (Duff: Vol. 1, 1.22).

The addition of a low pass or bandpass filter between the transmit-receive pair should easily resolve the out-of-band (TIM, RIM and SIM) problems.

For co-channel or adjacent interference, time-sharing, pulse shaping, or signal-bysignal cancellers can be used to potentially obtain EMC. FIM interference is generally more difficult and can even be as severe as requiring expensive redesign of the transmitter. Intermodulation induced interference generally cannot be fixed. The solution is to continue operations at a limited performance level or to resolve with frequency management. (Wilson: 7, 24)

Changes in location may change the area of influence and/or the radiation patterns. Changes include separation distance, position and attitude, natural terrain

shielding and line-of-site masking. This change also changes the aerodynamics of the system and could cause additional problems. (Wilson: 7, 24)

The operations environment must be prioritized for EMI situations in which no current technology solution can resolve the problem. In the power case mentioned by Wilson and Jolly, additional regulation or filtering of the power may help to meet the performance requirements. (Wilson: 24, 26)

USAF Multimission Aircraft Research and Development

Several key drivers are being worked as the United States Air Force (USAF) plans for 2025. One of the overarching drivers is the quest for information dominance. Joint Pub 3-13 defines information dominance (superiority) as "the capability to collect, process, and disseminate an uninterrupted flow of information while exploiting or denying an adversary's ability to do the same" (Joint Pub 3-13: I-11). To obtain the goal of continuous and uninterrupted flow of information, air space and information operations must be integrated seamlessly and quickly. Information technology is the key to sifting through the potential overload of information to deliver "the right information to the right place at the right time" (Jumper: 57) by horizontally integrating manned, unmanned and space platform command and control, communications and computers and intelligence, surveillance, and reconnaissance (C4ISR) systems. (Jumper: 57, 59)

A prime opportunity has come about with the need to replace the 40-year-old tanker fleet. The Boeing KC-135Es are scheduled to be replaced by a common widebody aircraft. The common frame of choice has been declared the Beoing 767 and work is under way to modify the commercial-of-the-shelf aircraft to accommodate the tanker and

transport missions. In addition to the tanker/transport missions, the replacement for the aging fleet of E-3, E-8, RC-135 and C-130 aircraft theater-based command and control (C2) and intelligence, surveillance and reconnaissance (ISR) fleet is underway.

In addition to the current platform retirement needs, the requirement of horizontal integration of C4ISR assets to accomplish information dominance over the battlefield can begin to become a reality with this widebody concept. The commander in charge of the battlefield must be provided all possible information in a timely manner in order to make the most accurate decision. This venture will almost resemble an air-based air operations center (AOC). The commander will have a complete air and ground battle management view to control the theater assets.

The common widebody aircraft integration referred to as MC2A or MMA will be an attempt to seamlessly incorporate current stove-piped theater C2ISR assets into a single cohesive unit. This is a fundamental change in the current acquisition process. It is proposed that the MMA be out-fitted to combine some or all the functions of the existing AWACS, JSTARS, RIVET JOINT, COMPASS CALL, and ABCCC platforms. It would also have links to other manned or unmanned ISR aircraft, as well as satellites. "The end result of this amalgamation of sensors, communications, and battle management elements will be the horizontal integration of surface, air and space-borne sensing and communications elements known collectively as the multi-sensor command and control constellation (MC2C) (Behler: 1)."

The USAF has established five integrated product teams (IPTs) to investigate the MMA development: 1) Concept of Operations (CONOPs) and Requirements IPT, 2) Threat and Scenario IPT, 3) Technology, System Concepts and Classified Systems IPT,

4) Modeling and Simulation IPT, and 5) the Acquisition Strategy IPT. These teams are comprised of members from Air Combat Command, Air Force Material Command, Air Force Space Command, and Air Mobility Command. The MC2A and MC2C concepts are highly praised and supported by the Air Force Chief of Staff General John Jumper and the Secretary of the Air Force Dr. James Roche (Paone, Roche).

In order to get the C2MA into the warfighters hands quickly, a spiral development approach has been chosen. The first spiral will consist of the Multi-Platform Radar Insertion Technology Program (MP-RITP) radar (JSTAR-like capabilities) incorporated with a battle management suite. The battle management suite will allow "cruise missile defense, control of unmanned aerial vehicles and time critical targeting" (Tuttle). The second spiral will be the incorporation of similar AWACS AMTI and C2 system capabilities. The passive remote sensors would be introduced in the final development phase (Tuttle, Fulghum: July 2002).

A report discussing the analysis of alternatives (AoA) for the MMA concept has been reported in an October 2002 study. Global Security summarized the "Alternatives for Joint Multi-Mission Aircraft" report as in Table 5 below:

Analysis Parameters Alternative	Estimated Cost	Number of Aircraft Required	Comments
Single Aircraft	\$189 Billion	176	Most costly and risky Estimated to take 3-5 years longer to field
Single A/C without signals- gathering capability	\$132 Billion	144	Same problems as single aircraft
Joint SIGINT program	\$23 Billion	32	AF must commit to larger 767 aircraft
Common Airframe	\$111 Billion	191	Could force Navy to buy bigger plane than needed

Table 5. Analysis of Alternatives

Several issues have been identified as key drivers for the integrated aircraft. David Fulghum discusses some of these decision variables in Aviation Week and Space Technology as (Fulghum: July 2002):

- The antenna location, number, and combination electromagnetic effects are not fully understood and must be accomplished before work begins on the GMTI radar.
- Aircraft aerodynamics being influence by top and bottom fuselage drag.
- Electrical power requirements. Will the current generators provide enough power (640kw) to support two major radar systems?
- Data fusion limitations.

In addition to the IPTs work has been in progress for establishing a final mission need statement (MNS) and CONOPS for both a MC2A 707 testbed and final MC2A. The MC2A 707 testbed also called Paul Revere has already accomplished its first flight during the Joint Expeditionary Force Experiment (JEFX) 2002. JEFX 2002 via the Paul Revere testbed is the means to solve some of the development concerns to include the previously mentioned drivers. The findings from the experiment confirm earlier integration concerns and potentially limiting results. Problems were identified with the operator workstations, unstable data links, classified network vulnerabilities, interference problems, burnt cards and wires, aircraft blockage and multiple formats between C2 and intelligence assets. The overall drive for the horizontal integration was proven successful. Dynamic retasking of ISR and complete view of the battlefield drove timelines down from hours to minutes (Fulghum: September 2002).

In the end, the challenge of incorporating the radar systems has been proven too difficult based on the current technology. Stephen Trimble quotes the Deputy Director of Information Dominance for Air Force Acquisition, Bobby Smart as saying "interference, power and weight are three concerns...with today's technology, with today's engine performance, it's prudent to think about this in terms of two separate fleets" (Trimble: November 2002). The final result from JEFX 2002 is the development of two fleets of aircraft. One fleet consisting of the GMTI mission elements and the other fleet with the AMTI mission elements.

III. Methodology and Tools

The Hall's morphological box, SMAD, and PSARE process were briefly discussed in the previous chapter. In this chapter, the methodology and tools of these three processes will be discussed. The chapter will conclude with a description of the tailored process used for the MMA analysis.

Hall's Seven Steps

The problem definition can be grouped into two components: 1) the introduction, background and discussion of the problem and 2) the interrelated elements. The title, scenario, professional backgrounds of system developers, scope, actors, partitioning of elements into relevant components, and isolation of subjective elements make up the first group. The needs, alterable, constraints, societal sectors, and a description of the interactions amongst these elements are the interrelated products developed in the problem definition step.

The interrelationships are described using a self-interaction matrix and/or a crossinteraction matrix. In the self-interaction matrix each element within a product is evaluated. The relationships between two products can be described using a crossinteraction matrix. The level of interaction can also be annotated in the matrix using symbols. Figure 3 is an example of a need self-interaction matrix and a need-alterable cross-interaction matrix. A cross- and self-interaction matrix is usually generated for the needs, alterables, constraints and societal sectors as shown in Figure 4 to show linkages between the problem definition elements.



Figure 3. Examples of Self- and Cross-Interaction Matrices



Figure 4. Problem Definition Linkages (Sage: 68)

The value system design step generates an objectives hierarchy/object tree with the final node incorporating measures of effectiveness (MOEs) to quantify how well the architecture being studied meets the criteria of effectiveness. The objective tree is used to create an objective self-interaction matrix. The last step is to generate a crossinteraction matrix between the objectives and the objective measures.

Brainstorming of concepts, alternative architectures and system designs are created during the system synthesis step. The problem definition, value system and system synthesis interaction matrices are joined together for a whole system view (Figure 5)



Figure 5. Program Planning Linkages (Sage: 74)

The fourth step, system analysis utilizes the previous three steps to model and assess the consequences of a given alternative architecture. The optimization step finds the best system given the value system design and constraints based steps 1-4. Once the alternative architectures have been evaluated based on the optimized value system design, an architecture is chosen to proceed with. This occurs during the decision-making step followed by the implementation of the next phase.

SMAD

The SMAD methodology is very similar to Hall's and therefore only the requirements definition will be discussed as it pertains to the MMA methodology. The requirements baseline development (steps 10 and 11) begins by identifying the customer and user of the product, prioritization of these customer's needs, and identification of internal and external constraints on the system. A tool called "Quality Function Deployment (QFD)" (Wertz: 78) is then used to evaluate the needs and the corresponding technical attributes. This QFD process evaluates the attribute and function development. Figure 6 is an outline of the 'House of Quality'.



Figure 6. House of Quality Structure

The house of quality structure is similar to the interaction matrices described in the Hall's methods and tools. The correlations or conflicts triangle is a representation of the inter-relationships of the 'Hows' whereas the relationships matrix is a crossinteraction of the 'Whats' and 'Hows'. To establish priorities of the 'Whats' and to define the trade space, the 'Whats' are multiplied by a weighting factor and the 'How' columns are then summed up. This evaluation helps to determine where additional analysis should be accomplished.

The functional requirements are then established and decomposed along with the flow. The functional requirements are converted into technical characteristics. Quantifiable requirements are established based on the above steps. Next, block diagrams are used to express a single architecture's interfaces and relationships. These functional requirements are decomposed into lower levels based on the predefined architecture (Wertz: 93).

HHP

HHP is the methodology of the PSARE process and in essence the concurrent development of the architecture, essential requirements, and enhanced requirements system specification models. Figure 7 is a generic view of this concurrent methodology.

The total system life cycle (Figure 1) relates to the system specification models (Figure 7) in the following ways:

- Essential Requirements Model and Enhancing and Deriving Requirements Model correspond to Required Capabilities Analysis
- Architecture Model corresponds to the Architectural Analysis

The HHP method begins with the external stakeholder needs being assigned to an architectural model or passed through for requirement decomposition. Process, control, time and module specifications are developed along with a dictionary to trace architecture-to-external-requirements. From the decomposition of the architecture, a requirement-to-requirement trace is generated to record process or dictionary parent/child relationships. This requirement-to-requirement traceability matrix is updated as the decomposed requirements are detailed or further derived. Once the requirements have been derived to their lowest level, the requirements are enhanced and allocation of the architecture elements to the requirements is accomplished producing an architecture-requirements traceability matrix or architecture dictionary. The architecture components are assigned using superbubbles that are drawn around the respective requirements. These assigned superbubble architecture modules are then decomposed into finer detail creating an architecture-to-architecture traceability matrix.



Figure 7. System Specification Models (Hatley: 191)

In creating the traceability matrices, the incoming requirements can be used to easily check for completeness and design criteria satisfaction. The traceability matrices also allow for history compilation and validation to justify its existence and allow for impact analyses for change impacts at a later time. It is important to note that this methodology has no beginning or end and can be started at any point. Additionally, only the modules, diagrams and specifications that make sense to be completed are accomplished.



The development models can be summarized in the following manner:

Figure 8. Development Models Summary (Hatley: 73)

The context diagrams are the baseline view of how the system interacts with its environment. The flow diagrams are the hierarchal representations of the components within the system. As discussed above, the specification and dictionary elements are derived during the flow diagram developments. The flow diagrams and specification models occur as many times as is necessary to decompose the system to its lowest detail required for development, whereas, the context and dictionary inputs have only one occurrence.

HHP successively defines lower-level functional and performance requirements using the process modules. The process module in the requirements model defines the functional requirements. The process module is a layered set of data flow diagrams (DFDs) with a data context diagram (DCD) at the highest level, and a process specification (PSPEC) at the lowest level of each vertical thread (also includes a time specification (TSPEC) and a requirements dictionary (RD)). The requirements are traced back to the physical performance constraints/capabilities (which drive requirements) in the architecture model (data flow diagram (DFDs) by using superbubbles.

The architecture (physical) models handle the functional interfaces and architecture. The physical architecture is described using the flow module (architecture flow context diagram (AFCD) and the subsequent architecture flow diagrams (AFDs)) while the functional interfaces of the architecture are handled using the interconnect modules (architecture interconnect context diagram (AICD) and its subsequent architecture interconnect diagrams (AIDs)). Note that in the HHP methods, an interconnect consist of two or more interfaces.

Functional and performance requirements track with higher-level requirements. The modules (both under the requirements (functional) and architecture (physical performance) models) use a naming convention using unique singular nouns and numbers. Each child diagram maintains the noun and numerical identifier from its parent diagram. The grandchild maintains the naming convention from both its parent and

grandparent. Each new layer's (child diagram) numbers must comprise the diagram number (from the parent) appended by one additional number.

System requirements are allocated and defined in sufficient detail to provide design and verification criteria to support the integrated system design. The new layers of the data flow diagrams are created and the process specification completed to a point that the developed and defined system can be handed over to the developer. If the new system (generated from a top-down approach) is being integrated into an existing or legacy system (generated from a bottom-up approach), concurrent development and trade-off studies are needed. The HHP method suggests creating a sample analyzer module using the existing sampling module and comparing "the top-level model and the System Analyzer, the remainder of the system-level architecture, the allocation of requirements to the remaining architecture modules, and any further decompositions that are needed of those modules (Hatley: 347)." As can be seen by Chapter 11 of PSARE, the tracking of this integration can be followed using the enhanced requirements model. The enhanced requirements model elements are then allocated to the system using superbubbles.

System interface control requirements that are developed are fully documented. All requirements generated by the functional and physical architectures are document using specifications (process specifications (PSPECs), control specifications (CSPECs), timing specifications (TSPECs), architecture interconnect specifications (AIS), and others) and integrated dictionaries (requirements and architecture dictionaries).

The HHP method is a very detailed method for defining and decomposing the physical and functional components. However, the HHP method falls short when

comparing multiple alternatives. The value system design was used to evaluate multiple alternatives in the Hall's and SMAD methodologies.

The choice of the overall system architecture and the major technologies it will include could be a major part of system development in a new or complicated system. Hatley, Hruschka and Pirbhai suggest that there are three tools used "to make architectural and technological decisions...feasibility analyses, trade-off studies, and prototypes (Hatley: 202)" with detail increasing in the order stated. HHP method states that criteria to measure the various alternatives needs to be established in advance and weighted according to their relative importance.

"A trade-off study is the consideration of several potential architectures or designs to compare their pros and cons, and either to select one of them as the best candidate, or to look for other candidates. Trade-off study results are recorded in the rationale sections of AMSs and AISs (Hatley: 417-18)." AMSs generally "contain numerous references to trade-off studies, company and industry standards, other systems in the same family and other specifications (Hatley: 382)."

An email from Hatley suggests that a complete model must be developed for each alternative system. And the individual models compared. Additionally, "a tool that automates the methods can make populating the repository and checking its consistency much easier. Nevertheless, all the actual thinking, the problem-solving, the trade-off studies, and the myriad of other development activities must be done by you, the system developer (Hatley: 200)."

In considering the feasibility of two or more choices for a given entity, the alternatives could be listed as different attributes to the alternative. Once a decision is

made, the alternative attribute of choice could be annotated in the architecture module specification.

For the more detailed (full scale) comparison (trade-off study) of the alternative's cost, schedule, resource availability, and organizational politics, the design process would be completed on one (the most attractive, unique) alternative and then the next alternative with some change would be completed. These final results would be weighted based on criteria determined in advance. This process could become very overwhelming and rigorous with 2^n possible alternatives. It was suggested that the top two or three alternatives be chosen. These top alternatives would then be optimized based on the other alternatives.

A software tool called TURBOCASE has been designed based on the process and methodology of Hatley, Hruschka and Pirbhai. One of the greatest benefits to this graphical tool is its ability to perform consistency checks. These checks ensure that information going into and out of the modules are consistent and each module is traced back to a higher level reference. At the lowest level, the tool validates that all dictionary entries and specifications have been completed. The software is based on the unified modeling language (UML), structured analysis and structured design.

IV. Process Tailoring and EMI Results

Each of the methods discussed in Chapter 3 have their strengths and weaknesses. The Hall's methodology and SMAD methodology give great emphasis on the development of requirements and system definition. They both also include an analysis of alternatives. They do not, however, go into great detail on how to map requirements to architecture components. I do not feel the HHP methods fully address the build-up of the system definition and analysis of alternatives. The HHP methods seem to be more applicable once the up-front analysis has been performed. Once an architecture is defined, the HHP methods become the stronger method, as it is able to automatically check for consistency throughout the system model. Additionally, the HHP methods and tools allow for an automated process to track and map all of the system requirements to the architecture. For these reasons, the methods used in this study have used components and/or ideas from all three methods discussed.

Preliminary Analysis

The investigation of the multimission aircraft began with a preliminary group of twelve students comprised of logistics and maintenance operations, air and space operators, and acquisition, science and engineering backgrounds. The preliminary group brainstormed and researched the current platforms to develop two baseline hierarchies and value system designs using Hall's Seven Steps. In addition, a concept map (Appendix 1.2) was constructed to show relationships between key players, systems and operational considerations.

The objectives, needs, alterables, and constraints were analyzed in both a crossand self-interaction matrices similar to Appendix 1.3. However, only the interactions themselves were annotated, not the level of interaction.

These matrices were then used to develop a modified interface and flow models (Appendix 1.4) using techniques defined by Hatley, Hruschka and Pirbhai. The user interface model established the baseline user interface, input processing, output processing, main functions and the support functions. The system requirements and architecture model development suggested by HHP was used primarily to summarize and graphically track the multiple architecture developments. The previously defined user interface model was used to stimulate system specifications to iteratively generate a set of system requirements and architecture models. The interface model depicts key requirements and interactions within the MMA design. The process interface was the centerpiece or driving force behind the iterations. As each architecture was developed, the system requirements were enhanced and fed back into the interface. As the process continued, several architecture variations developed and are noted as sub-bullets in the systems architecture model shown in figure A.1.4. The system architectures defined by the group were:

- Baseline: The current standings of each mission without future improvements. Today's System.
- Legacy Improvements/ Standard Acquisition Process: Follows the traditional method followed by DOD in replacing aircraft. Under the Legacy concept, each weapon system will be replaced by a similar upgraded system. The degree of enhancements will be determined case by case by using inputs from the

commands and the System Program Office. Legacy replacement results in system architecture almost identical to that of today. This is classical "stove-piping," but given widely different schedules, budgets and technical risk, it remains a viable alternative.

- *One Tail Number (OTN)*: This would entail consolidation of multiple missions under a single airframe. This is the desired outcome from decision makers as it is expected to reduce life cycle cost and increase the ability to fuse data information, the original vision of the MMA.
- Different Tail Number (DTN): Each aircraft would consist of sets of compatible missions. For example tail number A1 may consist of Battle Management, C2 and IFF; tail number B2 may consist of C3CM, GMTI, IFF; tail number C3 may consist of C3CM and ISR; etc. Depending of the mission a tail number or set of tail numbers would be selected.
- *Receive-Transmit-Command (RTC):* The architecture consists of a suite of three types of aircraft missions. This concept centers on separating the three basic functions of systems described earlier into transmitting platforms, receiving and processing platforms, and separate command and control platforms. More than three aircraft could be used in the architecture but would be limited to one of the three primary missions.
- *Sensor Craft*: This is a long dwelling, real estate unlimited aircraft that could accomplish all potential missions under one aircraft.

• *Modular*: The aircraft would have a compartment or module that could be inserted based on the mission. Each module would be outfitted with different hardware and software specific to a missions needs.

MMA Thesis Team

The MMA thesis team consisted of a group of three students including Lt Nevin Coskuner, TUAF, Lt Ahmet Kahraman, TUAF, and myself. The MMA thesis team reinvestigated, compiled and developed a new and complete baseline including a systems definition consisting of key players, stakeholders, needs, alterables and constraints. The interaction matrix was reinvestigated based on these system definitions to visually show levels of cross-interactions.

The interaction matrix found in Appendix 1.3 was a key element in building the system synthesis architecture as it identified where special or in-depth research was needed to be accomplished. To logically assign levels of interaction, the designated strengths, high, medium, and low, were assigned numerical values of 9, 5 and 1, respectively. This is similar to the SMAD Quality Function Deployment in that each element value was totaled based on its interaction among the other elements. For each group (objective, alterable, constraint and need), the elements were arranged in order based on this total and natural group interaction levels were established. The cross-interactions have been summarized and categorized by level in Table 1. The analysis of the interaction-matrix determined the system variables that drove the design the most or at the "highest level." Other interaction levels were addressed as needed.

The MMA thesis team determined three key areas for further investigation from the cross-interactions of the system definition constraints. These areas consisted of the

operations environment, system compatibility, and payload limitations based on airframe limits. The technology availability and development time constraints were of higher interaction, however, it was felt among the group that the airframe limits and system compatibility would bring out some of these constraint details and in essence be addressed.

A value system design was also compiled based on the original group analysis using Hall's Seven Steps. Reduced life cycle costs and increased system value through measures of mission utility, mission integration and compatibility, and minimal risk are the primary objectives considered. The overall need was to ensure that every mission currently being served by this fleet will not only continue but also enhance a theater's ability to perform time critical targeting (TCT). Therefore, the MMA layered model was designed based on these goals and objectives and is as follows:



Figure 9. MMA Layered Model

In a review of the group alternative architectures, two main variations existed: 1) One Tail Number (all missions within a single aircraft) and 2) Different Tail Numbers (tail numbers representing the different sets of sensors within a particular aircraft). These two architectures were investigated in the individual study areas. Several DTN alternative architectures were generated based on the different combinations of the current aircraft functionalities. The OTN and DTN alternative architectures were evaluated using the operational scenarios discussed by Coskuner along with the sensor compatibility and payload design results. Table 6 is a summary of the architectures along with their alternative title.

ARCHITECTURE TYPE	ARCHITECTURE ALTERNATIVE	ALTERNATIVE TITLE	ON BOARD A/C
One Tail			AWACS
			JSTARS
	OTN	OTN	Rivet JOINT
			C.CALL
			ABCCC
		DTN11	AWACS
		DINII	JSTARS
	DTN1		Rivet JOINT
Different Tail Numbers		DTN12	C.CALL
			ABCCC
		DTN21	AWACS
		D11121	ABCCC
	DTN2		JSTARS
		DTN22	Rivet JOINT
			C.CALL
		DTN31	AWACS
			Rivet JOINT
	DTN3		JSTARS
		DTN32	C.CALL
			ABCCC
		DTN/1	AWACS
		DINHI	C.CALL
	DTN4		JSTARS
		DTN42	Rivet JOINT
			ABCCC

Table 6. OTN and DTN Alternative Architectures

Payload Integration as it Pertains to Electromagnetics

An EMC/I model was generated based on abstracted/detailed relationship modeling. In this type of relationship modeling, the downward usage adds detail or specializations, whereas the upward usage is used to generalize or abstract its subordinates. This type of model is generally used for requirements modeling to help reduce the complexity of the system at hand. Figure 10 shows this hierarchal decomposition.

As stated earlier, the antenna-antenna radiated intrasystem EMC will be investigated. Specifically, the transmit and receive spectrum power densities will be evaluated.



Figure 10. EMC/I Layered Model

"Short Form EMI Prediction"

Input Parameters

The preliminary EMC analysis tool discussed in Chapter 2⁵ was used to evaluate the transmit-receive antenna combinations. The quick look, amplitude, and frequency analysis sections were performed to determine feasibility. As applicable, the FIM, TIM, RIM, and SIM cases were tested. Advancement of the antenna-antenna pair into the next level was based on failure to meet the baseline IM requirement. As each level of analysis was performed, 90 percent of the non-interfering situations should be removed.

The input parameters are listed in Appendix 3.3, Table A.3.2 for each transmitreceiver combination. The transmitter-receiver combination looked at for EMI analysis include: 1) GMTI-AMTI, 2) AMTI-GMTI, 3) IFF-GMTI, 4) GMTI-IFF, 5) AMTI-high band⁶ (HB), 6) AMTI-low band⁷ (LB), 7) AMTI-SHF, 8) GMTI-HB, 9) GMTI-LB, and 10) GMTI-SHF. The transmit input parameters required for the analysis include:

- Frequency: The mean of the working band i.e. X-Band is 10000 MHz for GMTI.
- Power Output: An example in *A Handbook Series on Electromagnetic Interference and Compatibility (Vol 1), Fundamentals of Electromagnetic Compatibility* assigned a power level of 200dBm to a similar radar system. This value of 200dBm is used for all cases.
- Antenna Gain: Based on the standard radiation characteristics for a given type of antenna. Aperture or array antennas are generally 25-60

⁵ "Short Form EMI Prediction" tool developed by Don White Consultants in 1972. *The Electromagnetic Compatibility Handbook* by Violette, White, and Violette and *The Handbook Series on Electromagnetic Interference and Compatibility* (Vol. 7) by Duff both discuss this form in detail.

⁶ High frequency (HF) and high band (HB) are used interchangeably in this paper.

⁷ Low frequency (LF) and low band (LB) are used interchangeably in this paper.

dB/Isotrope (Duff: Vol. 1, 3.32). The IFF was chosen based on several IFF systems in *Janes's C4I Systems*. The gain could also be calculated as a function of the effective aperture and the frequency, but since the numbers are erroneous an estimate was given.

- Bandwidth: Except for IFF, the bandwidth was generically chosen.
 During the analysis, the bandwidth was adjusted to force compatibility⁸.
 The receiver input parameter information was based on:
 - Frequency, antenna gain, and bandwidth: The same as the transmitter rational.
 - Intermediate Frequency and Fundamental Sensitivity: Except for IFF, these values are based on examples used in *A Handbook Series on Electromagnetic Interference and Compatibility (Vol 1), Fundamentals of Electromagnetic Compatibility.*
 - Local Oscillator: The Frequency plus the intermediate frequency.

The last portion of the parameter inputs covers the placement and distance between the sensors on the aircraft. The height was based on a ratio estimate using the Boeing 767 average radius of 5.4 meters. The distance was then calculated using right spherical triangles. The placement of the LB, HB and SHF equipment was tested at two locations: 1) just above either side of the GMTI sensor and 2) 180 degrees below the AMTI radar. The placement of the AMTI and GMTI were established based on their current locations and can pictorial be seen in Figure 11.

⁸ This is discussed in the "Short Form EMI Prediction" Results section.



Figure 11. Schematic of Antenna Placement

<u>'Quick-Look' Analysis</u>

The short form model begins by setting the transmitter and receiver frequency limits bandwidths to 0.1 to 10 times the fundamental frequency. The maximum fundamental frequency separation is determined by 0.2 times the fundamental receive frequency. The calculated limits are then compared to determine SIM, RIM, TIM, and FIM. The minimum and maximum frequencies are tested for overlap. If overlap exists, the result will be positive for interference resulting in a yes response, whereas, no overlap corresponds to a negative (no) interference response. The limits and 'quick-look' analysis results can be found in Appendix 3.3, Tables A.3.3-3.6.

The sensor compatibility results of the preliminary short analysis showed that AMTI and GMTI combinations are expected to have interference due to the harmonic and spurious responses. This interference can be resolved by ensuring the bandwidth for the minimum transmit spurious frequency is greater than the maximum receive spurious frequency and the maximum transmit spurious frequency is less than the minimum receive spurious frequency. The GMTI and IFF combinations resulted in a compatibility issue as well. This is also the case with AMTI and IFF combinations. This is currently resolved on AWACS with time management. Therefore, GMTI and IFF antenna combination compatibility will be assumed to be manageable. Lastly, the AMTI and HB resulted in transmit harmonic and receive spurious interference. Unlike the AMTI and GMTI combinations, the AMTI and HB bandwidth cannot be adjusted to deconflict the wave patterns. The addition of RF filters for both the transmit and receive antenna and/or time management could be a potential solution. These results are summarized in Table 7. In this case, the no responses represent a probability that no interference exists, whereas, the yes represents a probability that interference will occur.

EMI Prediction Case	Antenna Combination					
	AMTI-RX GMTI-TX	AMTI-RX GMTI- TX*	GMTI-RX AMTI-TX	GMTI-RX AMTI-TX**	GMTI-RX IFF-TX	IFF-RX GMTI-TX
FIM	No	No	No	No	No	No
TIM	Yes	No	Yes	No	Yes	Yes
RIM	Yes	No	Yes	No	Yes	Yes
SIM	Yes	No	Yes	No	Yes	Yes
	AMTI-TX HB-RX	AMTI-TX LB-RX	AMTI-TX SHF-RX	GMTI-TX HB-RX	GMTI-TX LB-RX	GMTI-TX SHF-RX
FIM	No	No	No	No	No	No
TIM	No	No	No	No	No	No
RIM	No	No	No	No	No	No
SIM	Yes	No	No	No	No	No

Tuble 77 Entri Frederich Cube Repuits for Antenna Combinations	Table 7.	EMI	Prediction	Case	Results for	· Antenna	Combinations
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* Minimum transmit spurious frequency is greater than the maximum receive spurious frequency.

** Maximum transmit spurious frequency is less than the minimum receive spurious frequency.

Although the systems are more than likely digital systems, an analog short form for out of band interference was performed (Appendix 3.3, Tables A.3.11.a.-A.3.12.b). These results predicted no harmonic or spurious interference for the AMTI and GMTI, and GMTI and IFF antenna combinations. All of the AMTI and HB/LB/SHF, and GMTI and HB/LB/SHF resulted in harmonic interference predictions. Only the AMTI and SHF, and GMTI and SHF combinations are expected to have spurious interference. A more detailed analysis of the transmitter noise, third order intermodulation, receiver intermodulation, and transmitter intermodulation should be performed.

Amplitude and Frequency Culling

The transmit-receive antenna combinations resulting in a positive probability of interference were advanced to the amplitude level of analysis. The antenna gain, direction, propagation loss, transmit power, power available at receiver, and receiver

susceptibility (sensitivity) are used to calculate an IM for each surviving case. None of the cases passed the amplitude culling IM test and were advanced to the frequency culling. In the frequency culling analysis, the bandwidth is corrected. None of the cases passed the frequency culling IM test. At this point the analysis was stopped and the surviving cases were determined to be incompatible. The incompatible transmit-receive cases were determined to be AMTI-GMTI, GMTI-AMTI, GMTI-IFF, IFF-GMTI, and AMTI-HB. The analysis and results can be found in Appendix 3.3, Tables A3.7-A.3.10.b.

Impact to OTN and DTN Architectures

After the antenna combination compatibility analysis was completed, the OTN and DTN alternatives were analyzed and summarized in Table 8. For the combinations containing the Rivet Joint electronic counter measure for which no evaluation was performed, the overall compatibility was based on the assumption that time management and antenna direction could control the EMI. In addition, there is research currently being accomplished in which SIGINT (the HB, LB, SHF functions) and ECM are being combined into one physical component. This single module will focus the transmit or receive variable wavelength in a specific direction instead of omni-receive/transmit (Fulghum: 34).

ARCHITECTURE	ARCHITECTURE	ALTERNATIV E	A/C	SENSOR		
ТҮРЕ	ALTERNATIVE	TITLE	ON BOARD	COMPATIBILITY		
			AWACS	AMTI-GMTI	Yes [%]	
			JSTARS	GMTI-AMTI	Yes%	
One Tail	OTN	OTN	Rivet JOINT	AMTI-RJ GMTI-RJ	No ^{%%} Yes	
Number			C.CALL	C. Call No Eval	uation	
			ABCCC	Comm Assume Yes		
	O	TN Overall		No		
		DTN11	AWACS	AMTI-GMTI	Yes%	
			JSTARS	GMTI-AMTI	Yes%	
	DTN1		Rivet JOINT		Yes	
		DTN12	C.CALL	C. Call No Eval	uation	
			ABCCC	Comm Assume Yes		
	DT	TN1 Overall	Assume Yes*			
Different Tail Numbers	DTN2	DTN21	AWACS	AMTI-Comm	Assume Yes	
			ABCCC	Comm-AMTI	Assume Yes	
		DTN22	JSTARS	GMTI-HB/LB/SHF	Yes	
			Rivet JOINT	C. Call No Eval	uation	
	DTN2 Overall			Assume Yes*		
			AWACS	AMTI-HB	No	
		DTN31	Rivet JOINT	AMTI-LB/SHF	Yes	
	DTN3	DTN32	JSTARS	GMTI-Comm	Assume Yes	
			C.CALL	C. Call No Eval	uation	
			ABCCC	Comm-GMTI	Assume Yes	
	DTN 3 Overall			Assume Yes*		
	DTN4	DTN41	AWACS	Assume Yes*		
			C.CALL	C. Call No Eval	uation	
		DTN42	JSTARS	GMTI-HB/LB/SHF	Yes	
			Rivet JOINT	GMTI-Comm	Yes	
			ABCCC	Comm-GMTI	Assume Yes	
	DT	N 4 Overall	Assume Yes*			

Table 8. OTN and DTN Alternative EMC/I Summary

* Assume time management capability for ECM

% Assume time management capability for IFF and GMTI/AMTI. This is how compatibility is currently achieved for IFF and AMTI on AWACS.

%% Time management may be an option for compatibility. The AMTI-HB function is expected to have harmonic and spurious interference.

Analysis of Results

A summary of the MMA results including the compatibility and the payload limitations is provided in Table 9. The OTN architecture has been determined to be infeasible due to limits imposed by both EMC and power supply. As discussed earlier, time and directional control techniques may be possible solutions to overcome EMI. Power supply management may help to overcome the supply issue, however, this is probably not realistic due to the extreme overtasking of the supply. The aircraft system being used is a commercial off-the-shelf platform and therefore comes with a standard power supply unit. The addition of more power units (APUs) to supply the required energy draw is not a one-for-one trade and is not very efficient.

The power supply limitations limit all of the alternative DTN architectures. For additional information pertaining to the generation of the power limits and payload limitations in general, the reader is referred to Kahraman's thesis.

Table 8 breaks down the sensors into individual antenna-antenna evaluations. The overall system interference evaluation assumes that all the systems would be used at the same time. .Lt. Coskuner's scenario evaluations may actually show that the systems can reside on the same platform and still function based on whether the specific system set will be operational at the same time.
ARCHITECTUR E ALTERNATIVE	ALTERNATIVE TITLE	ON BOARD A/C	MAX WIEGHT	CREW #	WEIGHT	POWER	REFUELS REQUIRED (SEA LEVEL)**	REFUELS REQUIRED (FROM 8000 FT)**	Endurance (hr)	SENSOR COMPAT	IBILITY
		AWACS						3	9	AMTI-GMTI	Yes%
		JSTARS								GMTI-AMTI	Yes%
OTN	OTN	RJ	?	Yes	?	No	2			AMTI-RJ GMTI-RJ	No%% Yes
		C.CALL								C. Call No Evaluation	
		ABCCC								Comm Assume Yes	
(OTN Overall									No	
	DTN11	AWACS	Var	Var	Var	No	2	2	12	AMTI-GMTI	Yes%
	DINII	JSTARS	res	res	res	INO	2			GMTI-AMTI	Yes%
DTN1		RJ						2	12		Yes
	DTN12	C.CALL	Yes	Yes	Yes	No	2			C. Call No Evaluation	
		ABCCC								Comm Assume Yes	
D	TN1 Overall		Yes	Yes	Yes	No				Assume Yes	5*
	DTN21	AWACS	Ves	Ves	Ves	Ves	2	2	12	AMTI-Comm	Assume Yes
	D11121	ABCCC	105	105	105	105	2			Comm-AMTI	Assume Yes
DTN2	DTU22	JSTARS	37	37	37	N	2	3	11	GMTI-HB/LB/SHF	Yes
	DIN22	RJ	r es	Yes	Y es	NO	2			C. Call Na Evaluation	
		C.CALL	V	V	V	NI.				C. Call No Evaluation	*
L	1 N2 Overall	AWACS	res	res	res	INO		2	12		No
	DTN31	RI	Yes	Yes	Yes	No	2	2	12	AMTI-LB/SHF	Yes
DTN3		JSTARS		-		-		3	11	GMTI-Comm	Assume Yes
	DTN32	C.CALL	Yes	Yes	Yes	No	2			C. Call No Evaluation	
		ABCCC								Comm-GMTI	Assume Yes
DTN 3 Overall		Yes	Yes	Yes	No				Assume Yes	¢	
	DTN41	AWACS	Yes	Yes	Yes	No	2	2	12	Assume Yes*	
		C.CALL					_		10	C. Call No Evaluation	
DTN4		JSTARS						2	12	GMTI-HB/LB/SHF	Yes
	DTN42	JOINT	Yes	Yes	Yes	No	2			GMTI-Comm	Assume Yes
		ABCCC								Comm-GMTI	Assume Yes
D	TN 4 Overall		Yes	Yes	Yes	No				Assume Yes	k

Table 9. MMA Feasibility Summary

Functional (Requirements) Design & Architecture (Physical) Design

An HHP system architecture model (physical model) and requirements model (functional model) were generated using TURBOCASE. The models can be found in Appendix 2.1 and Appendix 2.2. The dictionary is listed in Appendix 2.3. The output from the preliminary EMC/I analysis was incorporated into the HHP model via a data table.

These models are the start to a physical development activity. Therefore, they are not comprised of specifications and detailed requirements, as this was not the focus of the study.

VI. Conclusion & Recommendations

The start to a MMA system design has been accomplished using Hall's, SMAD, and HHP methodologies. Every attempt has been made to make the system as complete as possible. However, with every new set of eyes comes new views and in essence new inputs. A system architecture is never complete for this reason.

As stated earlier, there are basically two types of studies: feasibility studies and studies with a product implemented. A feasibility study highly focuses on needs, alterables and constraints to develop alternative architectures and recommendations for implementation. A detailed value system design is established. However, there is an overall lack of emphasis on system requirements.

With that stated this study did not concentrate heavily on the product to be implemented, and therefore, a focus on detailed requirements development and specifications was not included. The HHP methodology begins to address the product development but more analysis must be accomplished once an architecture has been decided. In addition, the HHP model focuses on the OTN. The OTN is broken down into great detail, whereas, the DTN is only broken down into the mission objective combinations.

The study has shown the OTN architecture, the most desired by the customer, to be infeasible due to compatibility issues and power limitations. This is in agreement with results from the Paul Revere-MC2A testbed performance in JEFX-2002. The final result from JEFX 2002 recommended the development of two fleets of aircraft based on "interference, power [and] weight" (Trimble). One fleet consisting of the GMTI mission elements and the other fleet with the AMTI mission elements.

The EMC analysis performed was extremely limited. The true values were not used and the analysis technique was not as complete as some of the automated tools. Additionally, it was assumed that the communication systems would work in the new environment since the same communications systems work in the current environment. This is probably a poor assumption to make since the coupling of the multiple system waves could actually cause interference due to the intermodulation, harmonic or spurious frequencies. It is therefore, recommended that a spectrum analyzer be used to determine the true spurious and harmonic frequencies of the systems and an automated tool be used.

The last point of contention is that this study was accomplished without the continued input of the customer. This input is a vital part of a complete and accurate system design.

It is recommended that:

- The true values should be inputted into an automated analysis tool to obtain actual EMI results.
- The power estimates be iterated based on some known technology advancements. An example of this is the computer monitor. The 1988 equations used to compute power and weight for the computer monitors would be a significant difference when compared to a more compact and energy efficient flat-screen monitor. For an estimated 60 console work area, this could prove to be a significant difference.
- The operations evaluations should consider if all systems are required to function at the same time. Perhaps all of the systems can be installed onto

one aircraft with an accepted limitation that only certain systems could operate at the same time.

VII. Appendix

Appendix 1.1 System Definitions.

Key Players and Stake Holders

The MMA players include the decision makers (ACC, AMC, AFSOC, CINCs), owners/operators (Air Staff, Nav Air, Army) and stakeholders (theater commanders, fighters, bombers, combat search and rescue, support aircraft, etc.). The technical actors include the Boeing Company, Raytheon Corporation, and Northrop Grumman and other companies. Some of the necessary disciplines of the feasibility study members include physics (electromagnetic), logistics, operations, acquisition, and engineering (sensor, transistor, receivers, aeronautics, systems.)

Need: What the customer wants.

Continuous Operations: All weather, 24 hour/7 days a week.

Dissemination and Transmission: Any emission leaving the aircraft such as outbound communication of others and active remote sensing.

Command, Control and Communication Counter Measures (C3CM): The reduction or elimination of an adversary's use of their C3 components.

ISR Processing and Exploitation: The manipulation and data extraction of the collection data.

Receiving and ISR Collection: Inward communication from others and passive or active gathering of remote transmissions for intelligence data.

Air and Ground Battle Management (BM): The management and tracking of air and ground assets and adversaries.

Air and Ground Command and Control (C2)

Longterm Compatibility: It is desired to not only have this system meet the needs of today but also be designed to easily integrate future technologies.

Joint Service Interoperability: In today's environment, it is becoming more and more important to be able to leverage off of and communicate with sister services. <u>Alterable: What is proposed to be varied.</u>

System Architecture: Although a single aircraft is ideal and highly desired, a modular platform may need to be used if the functions of some of the current missions are incompatible. Therefore, the system architecture will select the airframe based on all of the mission components being consolidated into one permanent platform or into a set of modular platforms.

CONOPs: How a system is employed affects the multi-mission/system compatibility because having multiple missions also means having multiple interest, desires and goals. For example, the theater commander may see it necessary to collect ISR information in one location but be out of range for the C3CM mission. A fully outlined training, techniques and procedures (TTP) manual will need to be developed.

Mission Requirements: The missions must all perform together. Tasking, processing, exploitation and dissemination (TPED) will have to be investigated amongst the missions for system hardware and software overlap, independence, and interference. The space, weight and power requirements for the missions will also need to be evaluated.

Future Politics/Players/Conflicts/Demands: Each of these future aspects could drive the design and development of the MMA in a completely new vector giving way to a new set of requirements.

Constraints: What is held fixed.

Classification of the System: Each mission aircraft currently consist of, works at and reports at different levels of security. Bringing these different levels together and meeting security requirements may increase the difficulty in obtaining the overall integration.

Government Requirements and Policies: International and National level policies and regulations may restrict and even drive some of the decision variables.

Safety: Crew, data information and technology, and the aircraft safety will play a role in limiting the operations area and the optimal architecture. The higher the number of people onboard increases the safety concerns and could limit how close to a conflict the aircraft could safely fly.

Development Time: If the technology is not in already in place, it could extend the time required to develop an operational aircraft. If the development takes to long, a new proposed enemy/conflict could impact the requirements and the current design become infeasible.

Operations Environment: Trained personnel, aircraft/human survivability, friendly and hostile electromagnetic environment, and the overall mission performance will limit the ability of the MMA program.

Logistics Supportability: Transportation, manpower, supply, environmental impacts, and rapid return to service will all constrain the logistics and maintenance of an operational aircraft.

Technology Availability: In order to consolidate missions that currently require their own airframe, technology must be in place to minimize the real estate needs and

architecture systems of the missions. Newly designed transmitters and receivers will need to be designed that can handle the multi-missions.

Airframe Limits: Each airframe has its space, weight, range/endurance limits. The airframe must be able to manage the real estate and loiter requirements of the consolidated missions.

System compatibility: Will all of the different missions be able to work together? Will the C3CM mission prevent the C2 and communications and ISR collection missions form occurring? Therefore, the electromagnetic interference between transmitters and the interference between active and passive sensors will need to be investigated. Standardization, interoperability, and system supportability will all need to be considered.

Funding: The decision makers have not yet established the MMA funding level. Therefore, the life cycle cost approach will be to minimize the overall cost for the mission consolidation. This cost will be based on individual mission system requirements, modification cost of currently existing commercial airframe or development of a new airframe, open architecture cost and the consolidation of the missions into this architecture, a consolidated communications architecture, consolidated radar systems, console computer cost, size of aircrew required to perform the multimissions, processing (on-board or ground), etc. Additionally, if the MMA proves to be too costly and the cost outweighs the benefit then the aging fleet could age even longer.





Figure A.1.1. Relations Concept Map

The above context map identifies the interactions and relationships between the key players and the system elements. This map helps to define the interactions matrix in Appendix 1.3.

Appendix 1.3. MMA System Interaction Matrix.



Objectives

Х

1

Χ

Χ

Х

Х

Χ

1

Х х

Х х

Х

Х

Х

Х

1

х х Х х Constraints

Alterables

Interaction Matrices X / X X Air C2 (N11) X / Ground C2 (N10) X X / / Air BM (N9) X / Ground BM (N8) X ISR Coll.& Rec. Mis. (N7) ISR Process.& Expl. (N6) X / Dissem. & Transm. (N4) Joint Service Interoperability (N3) All-W eather Capability (24/7) (N2) Longterm Compatibility (N1)

Figure A.1.2. MMA System Interaction Matrix

Appendix 1.4. Generation of Alternative Architecture Concepts

User Interface Model

USER INTERFACE									
 Processing, Exploitation & Dissemination (PED) Ground Station Interaction Communications between Aircraft Aircrew Consoles 									
INPUT PROCESSING	MAIN FUNCTION	OUTPUT PROCESSING							
 Decision Maker Input Tasking Information from others IFF Signals Other Services Validation from other sources 	 Overall Command & Control Air/Ground Battle Management Sensors Collection – ISR C3CM SUPPORT Tankers Communications Relay Logistics & Maintenance 	 MASINT, SIGINT, IMINT, GMTI Decision Maker Output - Warfighter Direction 							

Figure A.1.3. User Interface Model

System Requirements and Architecture Model Development



Figure A.1.4. Systems Requirements and Architecture Model Development





Figure A.2.1. C/DFD-1: Context Diagram



Figure A.2.2. C/DFD: Perform C3CMISR



Figure A.2.3. C/DFD: Talk to Outside Sources



Figure A.2.4. C/DFD: Perform CM Task



Figure A.2.5. C/DFD: Command and Control AOI



Figure A.2.6. C/DFD: Perform ISR Tasks



Figure A.2.7. C/DFD: Check EMI

Input					Process				
AMTI Radar	GMTI Radar	SIGINT/ELINT Listen	Broadband Noise	Perform AMTI	Perform GMTI	Perform SIGINT/ELINT	Perform CM Task	Intersystem Compatibility	
On		Off		1				Yes	
Off	On	Off	,		1			Yes	
Off		On	Off			1		Yes	
Off			On				1	Yes	
On		Off		1	1			Yes	
On	Off	On	Off	1		1		No	
Off	On	On	Off		1	1		Yes	
On		· .	Off	1	1	1		No	
On	Off		On	1			1	No	
Off	On	Off	On		1		1	No	
Off		On				1	1	Yes	
On				1	1	1	1	No	

Figure A.2.8. Data Table: Check EMI



Figure A.2.9. C/DFD: Perform Tasks



Appendix 2.2. Physical Architecture Model

Figure A.2.10. AFD: Perform C3CMISR







Figure A.2.12. AFD: OTN



Figure A.2.13. AFD: Sensors



Figure A.2.14. AFD: AMTI Antenna



Figure A.2.15. AFD: GMTI Antenna



Figure A.2.16. AFD: IFF Antenna



Figure A.2.17. AFD: Battle Management



Figure A.2.18. AFD: DTN





Figure A.2.19. EC/DFD Enhanced Perform C3CMISR



Figure A.2.20. EC/DFD Enhanced Perform C3CMISR with Supperbubbles

Name	Definition	Туре	Source	Destination
AMTI Radar	*Air Moving Target Indicator Transmit & Receive Active IFF Included	Data/ Control	Sensors	C2 Management
	*DOMAIN: [On Off]		AMTI	C2
			Receive	Management
			MMA	Processing Exploitation Disseminatio n
			C2 Management	onboard processor and exploitation consoles
AOI Needs	*Needs requested by the area	Data	Theater Input	C2 Assets
	commander and allied forces*		Theater Input	C2 Management
			Theater Commander	MMA
Bomb-on-	*Mission tasked to "Kill"	Data	MMA	Operators
target	aircraft or other source to			
	destructive means upon a			
	foreign system*			
Broadband	*Broadband Noise from ECM.	Data/		
Noise	Noise that is spread throughout	Control		
	a large spectra. *DOMAIN: [On Off]			

Appendix 2.4 Architecture and Requirements Dictionary

Name	Definition	Туре	Source	Destination
CM Task	*Electronic counter measures	Data	C2	Sensors
	tasks		Management	
	*Battlemanagement (BM)			
			C2	C2 Assets
			Management	
			C2	UAV
			Management	
				a 1514
			A1r BM	Ground BM
			Ground BM	AIT BM

Name	Definition	Туре	Source	Destination
Exchange	*The exchange of information	Data	Theater Input	C2Management
Battle	from or about the battlespace			
Info	to include ground and air		C2	Theater Input
	information and data collected		Management	
	from the system or from other			
	sources*		BM	C2 Management
			C2	DM
			C2 Management	DIVI
			Wanagement	
			UAV	C2 Management
			Onboard processor and exploitation consoles	C2 Management
			ground station data input	C2Management
			C2 Management	C2 Assets
			Processing Exploitation Disseminatio n	MMA
			Ground Station Interaction	MMA
			Allie Ground Support	Allie Air Support
			Allie Air Support	Allie Ground
			Allie Ground Support MMA	Support MMA MMA

Name	Definition	Туре	Source	Destination
GMTI Radar	*Ground Moving Target	Data/	Sensors	C2 Management
	Indicator Radar Transmit and	Control		
	Receive Active		GMTI	C2 Management
	*DOMAIN: [On Off]		Receive	
				Drogozzina
			IVIIVIA	Frocessing
				Dissemination
				Dissemination
			C2	onboard
			Management	processor and
				exploitation
				consoles
IFF Radar Signature	*Identification of friend or foe radar signature*	Data	Sensors	C2 Management
8			IFF Receive	C2 Management
				-
			C2	onboard
			Management	processor and
				exploitation
				consoles
				Drogoging
			IVIIVIA	Frocessing
				Dissemination
Inspection	*Maintenance The aircraft is	Data		
and	grounded*	2		
Repair				
Intersystem	*Check for system	Data/		
Compatibility	compatibility. The ability for	Control		
	components to function as			
	expected without hindrance			
	*DOMAIN:			
	[Yes No "MAYBE"]	D		
ISR Database	*Database containing all	Data	C2	
	spectrum information about		Management	
	aircraft to include range		Sansors	C2 Management
	nower nower density		5015015	C2 Wanagement
	spectrum*		C2 Assets	C2 Management

Name	Definition	Туре	Source	Destination
ISR Task	*Intelligence, surveillance, and	Data	C2	Sensors
	reconnaissance task performed		Management	
	by the aircraft sensors or other			
	sensors outside the aircraft		C2	C2 Assets
	*DOMAIN: [True False]		Management	
			C^{2}	UTAV
			C2 Management	UAV
			Wanagement	
			Air BM	Ground BM
			Ground BM	Air BM
Local	*The task is performed locally	Data		
	or within the system*			
Max Range	*Maximum range of	Data		
	influence*	D		
Mission Plan	*The plan of attack developed	Data	C2	
	prior to conflict and updated as		Management	
	information exchanged		Sensors	C2 Management
	through		5015015	C2 Management
	linough		C2 Assets	C2 Management
No	TBD	Data		
noise from	*Noise generated from the	Data		
others	environment or outside			
	sources. Intersystem			
	compatibility issue*	_		
Power	*Power required to sustain the	Data		
D AOI	system*	D (
Prepare AOI	*Preparation for a future	Data		
	battlespace. Used to generate			
Refuel	*Tanker operations for in-air	Data	ΜΜΑ	Tankers
Operations	refueling Increases the total	Data		1 direct 5
~ perutions	operation time without			
	landing*			
Remote	*Operations occurring outside	Data		
	the system*			

Name	Definition	Туре	Source	Destination
Secure	*Secure communications to	Data	Communicati	MMA
Comms	inside or outside the system*		ons Relay	
				Commo Dolor
Consor	*Choice of grater function to	Data	MMA	Comms Relay
Sensor	*Choice of system function to	Data		
Choice				
	["AMTI" "GMTI" "SIGINIT/F			
	LINT" "EM" "A-GMTI"			
	"AMTI-SIG" "GMTI-			
	SIG" "GMTI-IFF" "SIG-			
	IFF" "A-GMTI-SIG-IFF"]			
SIGINT/	*SIGINT/ELINT Listen	Data/	C2	onboard
ELINT	Active	Control	Management	processor and
Listen	*DOMAIN: [On Off]			exploitation
				consoles
			Sensors	C2 Management
				р :
			MMA	Processing
				Dissemination
Spectrum	*Power density spectrum to be	Data		Dissemination
Settings	used by a	Dulu		
~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~~	function/operation/antenna*			
Theater	*Direction given to the theater	Data	Theater Input	
Direction	based on battlemangement			
	information*		Theater Input	C2 Management
			C2	Theater Input
			Management	
			C	
			C2	C2 Assets
			Management	
			Theater	MMA
			Commander	
			MMA	Theater
				Commander
			Commander	MMA
			Taskings	

Name	Definition	Туре	Source	Destination
Train	*Education and training of	Data		
Personnel	personnel onboard and			
	offboard the aircraft includes			
	operation of system, data			
	processing & exploitation,			
	report generation,			
	maintenance, etc.*		<b>C2</b>	
Update AOI	*Updates to the area of interest	Data	C2 Managament	
	based on battlespace		Management	
	linoimation		C2	C2 Assets
			Management	02 / 155015
			management	
			C2	BM
			Management	
			C2 Assets	C2 Management
			BM	C2 Management
			G	<b>G2</b> ) (
			Sensors	C2 Management
				Draggad
			MIMA	Intelligence
				Data
				Duiu
			IPB	MMA
			MMA	IPB
Vehicle Task	*tasks given to other aircraft	Data	C2	C2 Assets
	or systems in the environment.		Management	
	UAVs are an example.*			
			MMA	Operators
			Air BM	Ground BM
			Ground RM	Air BM

# Appendix 3.1. Automated EMC/I Program Analysis (Wilson: 12)

			Modeling Attrib	utes		
Program Name (Developer)	Coupling Path Model	Antenna Model	Mismatch Considerations	Emission Spectrum	Receiver Susceptibility	General Comments on Program Suitability
<i>IEMCAP</i> Intrasystem Electromagnetic Compatibility Analysis Program (Rome Air Development Center, USAF)	Uses an infinite cylinder truncated at one end by a cone. Includes wing and off wing fuselage models.	Both low and high gain antenna models, but no frequency dependence.	Conceptually, off line computations on empirical data can be included at attenuation of an in- line filter model.	Both functional and non- functional spectral models for several types of modulation.	Arbitrary and MIL STD selectivity curves based on power threshold and integrated margin.	Would be a good program for computing RF system antenna coupled interferences if models for gain pattern vs. frequency and in-line filter/VSWR considerations are added.
SEMCAP Specification and Electromagnetic Compatibility Analysis Program (TRW Systems Group)	Off-line computation or empirical data must be supplied. Only field transfer functions are available.	Off-line computations or empirical data on both antenna-to- field and field-to- antenna transfer function.	Conceptually, could be included in a receptor filter card using data from separate off-line analysis.	Limited spectral types for RF systems.	Limited definition of receiver selectivity, uses integrated voltage referenced to a threshold for susceptibility criteria.	Excellent program for interference in cables wire routing, limited applicability to RF system antenna coupled interferences.
DECAL Design Communications Algorithm (Naval Ocean Systems Center, Navy)	No coupling path model is provided, instead, antenna deficiency (required isolation) is the primary output of the program.	Antenna models are not necessary since required antenna isolation is the output of the program.	No apparent considerations of mismatch is included. Coupler insertion losses, however, are included.	Detailed consideration of functional emission spectrum, spurious signals, and broadband transmitter noise.	Detailed consideration of spurious responses of receiver. Receiver impedance versus frequency is not included.	Good detailed program, however, the issue of antenna coupling is not addressed. Evidently, antenna coupling is left entirely to off-line analysis.

# Table A.3.1 Automated EMC/I Program Analysis (Wilson: 12)

			Modeling Attrib	outes		
Program Name (Developer)	Modeling Attributes	Program Name (Developer)	Modeling Attributes	Program Name (Developer)	Modeling Attributes	General Comments on Program Suitability
COSAM Co-Site Analysis Model (Electromagnetic Compatibility Analysis Center, DOD)	No coupling models other than free space, far field gain models.	Simple user supplied antenna gain models.	Transmission line impedance mismatch effects are included.	Detailed consideration of functional type of interference signal. No consideration of spurious emissions in models.	No model for receiver impedance or threshold effects at front-end (Note: This is not the program's purpose) Spurious models are included.	Primarily a probability of interference program using detailed models for (S+I)/N of receiver demodulation processes. Antenna gain models may not be entirely accurate.
AFMAP AWACS Frequency Analysis Management Program (Boeing Company)	Uses an infinite cylinder model. Does not include capabilities for off fuselage models.	Uses generic monopole antenna models and specialized, user developed, subroutines for other antennas. Can include frequency dependant gains.	Includes separate models for VSR and in-line filters. Data, however, must be known form off-line sources.	Detailed discrete voltage spectral models form data computed off-line.	General selectivity cure based on power threshold.	Good program for RF subsystem antenna coupled interference. AFMAP primarily acts as a data handler/computational aid and does not include specific models.
SCAPS Scattering and Propagation Simulator (General Electric)	Can account for interposed objects in a cluttered environment but offers on specific models.	Antennas are modeled as a coupled pair with scattering matrices. Assumes conjugate match conditions.	Capable of including all transmission system losses. No specific models are available, these must be developed on an individual basis.	General discrete emission spectra specified by user.	Represents receiver as frequency dependent reflection coefficient.	Very good systematic approach (scattering matrix) to antenna coupling including many affects. Models for specific interactions are, however, lacking.

### Appendix 3.2. Preliminary Electromagnetic Compatibility Analysis

### Input Parameters

Co-Site transmitters and 1			AMT GMT	TI-RX TI-TX	GMT AMT	TI-RX TI-TX	GMT IFF-T	TI-RX FX	IFF- GMT	RX T-TX			
TX Frequency			fT	MHz		10000		1500		1030	]	0000	l.
TX Power Output			РТ	dBm		200		200		200		200	l.
TX Antenna Gain			GT	dBm		60		60		25		60	1
TX Bandwidth			TBW	MHz		10		10		3		10	1
RX Frequency			fR	MHz		1500		10000	-	10000		1090	l.
RX Intermediate Frequency			IF	MHz		60		100		100	70		l.
RX Local Oscillator			LO	MHz		1560	60 1010		-	10100		1160	
RX Fundamental Sensitivity			PR	dBm		-100		-100		-100		-84	n
RX Antenna Gain			GR	dBm		60		60		60		25	l.
RX Bandwidth			RBW	MHz		10		10		10		8	l.
Coverage				nmi	250			200	0 256			200	
Distance between TX & RX			dTR	km	0.0	40889	0.0	40889	0.04	40889	0.04	10889	l.
Distance between TX & RX			dTR	miles	2.54E-02 2.		2.5	4E-02	2.54E-02		2.54E-02		l.
Length between sensors on aircraft				km	4.00E-0		4.0	0E-02 4.00		0E-02	2 4.00E-02		1
Height between sensors (radius of A/C)				km	5.40E-03		5.4	0E-03	5.40E-03		5.40E-03		l.
		/			•								n
Co-Site transmitters and receivers			AMTI-TZ HB-RX	X AMT LB-R	TI-TX X	AMT SHF-I	I-TX RX	GMT HB-R	I-TX X	GMT LB-R	I-TX X	GMT SHF-	I-TX RX
TX Frequency	fT	MHz	1500		1500		1500		0000	1	0000	10000	
TX Power Output	РТ	dBm	200		200		200		200	20		0 200	
TX Antenna Gain	GT	dBm	60		60		60		60		60		60
TX Bandwidth	TBW	MHz	10		10		10		10		10		10
RX Frequency	fR	MHz	1	17		1800	800000		17		0.08 18000		00000
RX Intermediate													
Frequency	IF	MHz	5		0.03	2000	2000000		5		0.03	2000000	
RX Local Oscillator	LO	MHz	22		0.11	3800	38000000		22		0.11 38000		0000
RX Fundamental	PR	dBm	_4	50	-50		-50		-50		-50		-50
RX Antenna Gain	GR	dBm		25 25	25		50		25		25		50
RX Bandwidth	RBW	MH7	2	25	100	3000	0000		25		100 3000		0000
Coverage	ICD W	nmi	13	30	130	5000	130		130		130	5000	130
Distance between TX &			1.		150		150		150		150		150
RX	dTR	km	0.04088	39 0.04	40889	0.04	0889	0.04	0889	0.040	0211	0.040	0211
Distance between TX & RX	dTR	miles	2.54E-(	)2 2.54	4E-02	2.54	E-02	2.54	E-02	2.49	9E-02	2.49	)E-02
Length between sensors on aircraft		km	4.00E-0	02 4.0	0E-02	4.00	)E-02	0.00	E+00	0.00	E+00	0.00	E+00
Height between sensors (radius of A/C)		km	8.28E-0	04 8.2	8E-04	8.28	3E-04	0.00	4572	0.00	4572	0.00	)4572

# Table A.3.2. System Parameters
### "Short Form Prediction" Tool

## 'Quick-Look' Analysis and Results

							Make BW
					Make BW		smaller
					smaller work?		work? Use
					Use defined BW		defined BW
				Generic	in Parameters	Generic	in Parameters
	Transmitter and Receiver Frequency Limits				IF fST(min) > fSR(max)		IF fST(max) < fSR(min)
				AMTI-RX GMTI-	AMTI-RX	GMTI-RX	GMTI-RX
Line	Parameter	Symbol	Equation - if needed	TX	GMTI-TX	AMTI-TX	AMTI-TX
1	TX fundamental Freq	f OT		10000	10000	1500	1500
	TX Minimum Spurious						
2	Freq	f ST (min)	0.1*f OT	1000	9995	150	1495
	TX Maximum Spurious						
3	Freq	f ST (max)	10*F OT	100000	10005	15000	1505
4	RX fundamental Freq	f OR		1500	1500	10000	10000
5	RX Minimum Spurious Freq	f SR (min)	0.1*f OR	150	1495	1000	9995
	RX Maximum Spurious						
6	Freq	f SR (max)	10*F OR	15000	1505	100000	10005
7	TX-RX Max Allowable Freq Separation for Fundamental FMI	delta f (max)	0 2*f OR	300	300	2000	2000
/			0.2 1010	500	500	2000	2000

#### Table A.3.3.a. AMTI and GMTI Bandwidth Definition

### Table A.3.3.b. AMTI and GMTI EMC 'Quick-Look' Results

Applicability of Four EMI					
SIM: TX Harmonic & RX Spurious					
	f ST (min) < f SR (max)	Yes	No	Yes	Yes
	f ST (max) > f SR (min)	Yes	Yes	Yes	No
If NO then there is not an EMI Problem = STOP	SIM EMI Problem?	Yes	No	Yes	No

<b>RIM</b> : TX Harmonic & RX Fundamental					
	f ST (min) < f OR	Yes	No	Yes	Yes
	f ST (max) > f OR	Yes	Yes	Yes	No
If NO then skip RIM; enter N/A on line 38	RIM EMI Problem?	Yes	No	Yes	No

TIM: TX Fundamental & RX Spurious					
	f OT < f SR (max)	Yes	No	Yes	Yes
	f OT > f SR (min)	Yes	Yes	Yes	No
If NO, skip TIM; enter N/A on line 38	TIM EMI Problem?	Yes	No	Yes	No
IF Both RIM & TIM were N/A, skip FIM and enter N/A on line 38					

<b>FIM</b> : TX Fundametnal &	RX Fundamental					
		abs(f OT - f OR) < delta f	N	N	N	N
		(max)	NO	No	No	No
If No, skip FIM; enter N/A on line 38		FIM EMI Problem?	No	No	No	No

_

					Make BW		Make BW
					smaller work?		smaller work?
					Use defined		Use defined
					BW in		BW in
				Generic	Parameters	Generic	Parameters
	Transmit	ter and Receiver	Frequency Limits		No		No
				GMTI-RX	GMTI-RX	IFF-RX	IFF-RX
Line	Parameter	Symbol	Equation - if needed	IFF-TX	IFF-TX	GMTI-TX	GMTI-TX
1	TX fundamental Freq	f OT		1030	1030	10000	10000
	TX Minimum Spurious						
2	Freq	f ST (min)	0.1*f OT	103	1028.5	1000	10000
	TX Maximum Spurious						
3	Freq	f ST (max)	10*F OT	10300	1031.5	100000	10000
4	RX fundamental Freq	f OR		10000	10000	1090	1090
	RX Minimum Spurious						
5	Freq	f SR (min)	0.1*f OR	1000	9995	109	1090
	RX Maximum Spurious						
6	Freq	f SR (max)	10*F OR	100000	10005	10900	1090
	TX-RX Max Allowable						
	Freq Separation for						
7	Fundamental EMI	delta f (max)	0.2*f OR	2000	2000	218	218

Table A.3.4.a. GMTI and IFF Bandwidth Definition

Applicability of Four EMI	Prediction Cases				
SIM: TX Harmonic & RX Spurious					
	f ST (min) < f SR (max)	Yes	Yes	Yes	Yes
	f ST (max) > f SR (min)	Yes	Yes	Yes	Yes
If NO then there is not an EMI Problem =					
STOP	SIM EMI Problem?	Yes	Yes	Yes	Yes

## Table A.3.4.b. GMTI and IFF EMC 'Quick-Look' Results

<b>RIM</b> : TX Harmonic & RX Fundamenta	l				
	f ST (min) < f OR	Yes	Yes	Yes	Yes
	f ST (max) > f OR	Yes	Yes	Yes	Yes
If NO then skip RIM; enter N/A on line 38	RIM EMI Problem?	Yes	Yes	Yes	Yes

TIM: TX Fundamental	& RX Spurious					
		f OT < f SR (max)	Yes	Yes	Yes	Yes
		f OT > f SR (min)	Yes	Yes	Yes	Yes
If NO, skip TIM; enter N/A on line 38		TIM EMI Problem?	Yes	Yes	Yes	Yes
IF Both RIM & TIM were N/A, skip FIM and enter N/A on line 38						

<b>FIM</b> : TX Fundametnal &	& RX Fundamental					
		abs(f OT - f OR) < delta f (max)	No	No	No	No
If No, skip FIM; enter N/A	on line 38	FIM EMI Problem?	No	No	No	No

					Make BW smaller		
				Generic	work?	Generic	Generic
	Transn	nitter and Receive	r Frequency Limits		No		
Line	Parameter	Symbol	Equation - if needed	AMTI-TX HB-RX	AMTI-TX HB-RX	AMTI-TX LB-RX	AMTI-TX SHF-RX
1	TX fundamental Freq	fOT		1500	1500	1500	1500
2	TX Minimum Spurious Freq	f ST (min)	0.1*f OT	150	1500	150	150
3	TX Maximum Spurious Freq	f ST (max)	10*F OT	15000	1500	15000	15000
4	RX fundamental Freq	f OR		17	17	0.08	18000000
5	RX Minimum Spurious Freq	f SR (min)	0.1*f OR	1.7	17	0.008	1800000
6	RX Maximum Spurious Freq	f SR (max)	10*F OR	170	17	0.8	180000000
7	TX-RX Max Allowable Freq Separation for Fundamental FMI	delta f (max)	0.2*f OR	3.4	3.4	0.016	3600000

 Table A.3.5.a.
 AMTI and HB, LB, and SHF Bandwidth Definition

Applicability of Four EM	Prediction Cases				
SIM: TX Harmonic & RX Spurious					
	f ST (min) < f SR (max)	Yes	Yes	No	Yes
	f ST (max) > f SR (min)	Yes	Yes	Yes	No
If NO then there is not an EMI Problem = STOP	SIM EMI Problem?	Yes	Yes	No	No

## Table A.3.5.b. AMTI and HB, LB, and SHF EMC 'Quick-Look' Results

<b>RIM</b> : TX Harmonic of	& RX Fundamental					
		f ST (min) < f OR	No	No	No	Yes
		f ST (max) > f OR	Yes	Yes	Yes	No
If NO then skip RIM; e	nter N/A on line 38	RIM EMI Problem?	No	No	No	No

TIM: TX Fundamen	tal & RX Spurious					
		f OT < f SR (max)	No	No	No	Yes
		f OT > f SR (min)	Yes	Yes	Yes	No
If NO, skip TIM; enter	N/A on line 38	TIM EMI Problem?	No	No	No	No
IF Both RIM & TIM w	ere N/A, skip FIM a	and enter N/A on line 38				

FIM: TX Fundametna	l & RX Fundamental					
		abs(f OT - f OR) < delta f (max)	No	No	No	No
If No, skip FIM; enter N/A on line 38		FIM EMI Problem?	No	No	No	No

	Transmitter and Receiver Frequency Limits												
				GMTI-TX	GMTI-TX	GMTI-TX							
Line	Parameter	Symbol	Equation - if needed	HB-RX	LB-RX	SHF-RX							
1	TX fundamental Freq	f OT			10000	10000							
	TX Minimum Spurious												
2	Freq	f ST (min)	0.1*f OT	1000	1000	1000							
	TX Maximum Spurious												
3	Freq	f ST (max)	10*F OT	100000	100000	100000							
4	RX fundamental Freq	f OR		17	0.08	18000000							
	RX Minimum Spurious												
5	Freq	f SR (min)	0.1*f OR	1.7	0.008	1800000							
	RX Maximum Spurious												
6	Freq	f SR (max)	10*F OR	170	0.8	18000000							
	TX-RX Max Allowable												
	Freq Separation for												
7	Fundamental EMI	delta f (max)	0.2*f OR	3.4	0.016	3600000							

# Table A.3.6 GMTI and HB, LB, and SHF EMC Bandwidth Definition and 'Quick-Look' Results

Applicability of Four EMI P				
SIM: TX Harmonic & RX Spurious				
	f ST (min) < f SR (max)	No	No	Yes
	f ST (max) > f SR (min)	Yes	Yes	No
If NO then there is not an EMI Problem				
= STOP	SIM EMI Problem?	No	No	No

RIM: TX Harmonic & RX	Fundamental				
		f ST (min) < f OR	No	No	Yes
		f ST (max) > f OR	Yes	Yes	No
If NO then skip RIM; enter	N/A on line				
38		RIM EMI Problem?	No	No	No

TIM: TX Fundamental &	RX Spurious				
		f OT < f SR (max)	No	No	Yes
	f		Yes	Yes	No
If NO, skip TIM; enter N/A	A on line 38	TIM EMI Problem?	No	No	No
IF Both RIM & TIM were					
38					

FIM: TX Fundametn	al & RX				
Fundamental					
		abs(fOT - fOR) < delta			
		f (max)	No	No	No
If No, skip FIM; enter N/A	on line 38	FIM EMI Problem?	No	No	No

	Amplitude and Frequend	AMTI-RX GMTI-TX				GMTI-RX AMTI-TX						
Line	Parameter	Symbol	Unit	Recommended Value/Equation If	FIM	TIM	RIM	SIM	FIM	TIM	RIM	SIM
Line		LITUDE C		Treeded	NA	11111	ICIIVI	SIN	NA	11111	KIIVI	51111
8	TX Power	PT (f OT)	dBm		1 17 1	200	XXX	XXX	1111	200	XXX	XXX
9	TX Spurios Power Output	Spurios Power Output PT (f ST) dBm				XXX	140	140	XXX	XXX	140	140
10	TX Antenna Gain in RX Direction	GTR (f)	dB	or 0dB		0	0	0		0	0	0
11	RX Antenna Gain in TX Direction GRT (f)		dB	or 0dB		0	0	0		0	0	0
12	Propagation Loss Using Freq No.	L			#1	#1	#4	#2	#1	#1	#4	#2
			MHz			10000	1500	1000		1500	10000	150
	Loss from Fig No. 2.7, (pg2 Duff (Function of freq,	2.36,Vol7) distance)	dB	all negative		82	64	61		64	82	44
13	Unintentional Power Available	PA (f)	dBm	PT + L	0	282	204	201	0	264	222	184
14	RX Fundamental Susceptibility	PR (f OR)	dBm	sensitivity		xxx	-100	XXX		XXX	-100	XXX
15	RX Spurious Suscept	PR (f SR)	dBm	PR(fOR) + 80dB	XXX	-20	XXX	-20	XXX	-20	XXX	-20
16	5 Preliminary EMI Prediction		dB	line 13-14 or 13-15	0	302	304	221	0	284	322	204
	IF EMI margin < -10 dB, EMI Highly Improbable = STOP IF EMI margin > -10 dB, Start Frequency Culling											

## Table A.3.7. AMTI and GMTI Amplitude Culling

					AMTI-RX GMTI-TX				GMTI-RX AMTI-TX				
Line	Parameter	Symbol	Unit	Recommended Value/Equation If Needed	FIM TIM RIM SI				FIM	TIM	RIM	SIM	
	FREQUENCY CULLING												
	Bandwidth Correcti	on											
17	TX PRF (if pulse)		pps			100	100	100		100	100	100	
18	TX Bandwidth	BT		2/PI * t if pulse; t=width		6.366	6.37	6.37		6.366	6.366	6.37	
19	RX Bandwidth	BR				10	10	10		10	10	10	
20	Adjustment	lines 17 to 19	dB	Use Fig 2.8/2.9		-92	-92	-92		-92	-92	-92	
21	Bandwidth Corrected	EMI Margin	dB	line 16+20	0	210	212	129	0	192	230	112	
	(BT+BR)/2				####	8.183	8.18	8.18	###	8.183	8.183	8.18	
	IF EMI MAR	LE = STOP											

## Table A.3.8.a. AMTI and GMTI Frequency Culling

					AMTI-RX GMTI-TX						GMTI-RX AMTI-TX			
Frequency Correction														
22 RX Local Oscillator Frequency	f LO	dBm				1560					10100			
23 RX Intermediate Frequency	f IF	dBm				60					100			
24 TX-RX Freq Separation:			delta f=abs((1)-(4)		XXX		XXX	XXX			XXX	XXX	XXX	
25  delta f > (BT+BR)/2			line (24), fig 2.10		XXX		XXX	XXX			XXX	XXX	XXX	
26 f OT/ f LO +/- f IF to nearest integer				XXX		6	XXX	XXX		XXX	0	XXX	XXX	
27 multiply lines (22) & (26)		MHz		XXX		9360	XXX	XXX		XXX	0	XXX	XXX	
28  delta f = abs((1)-(23)-(27))				XXX		732	XXX	XXX	•	XXX	1592	XXX	XXX	
28  delta f = abs((1)+(23)-(27))				XXX		548	XXX	XXX	1	XXX	1408	XXX	XXX	
29 select smaller delta f from (28)		MHz		XXX		548	XXX	XXX	•	XXX	1408	XXX	XXX	
30 delta f>(BT+BR)/2		dB	line (29), fig 2.10	XXX		-100	XXX	XXX	-	XXX	-100	XXX	XXX	
31 calculate f OR/f OT to nearest integer				XXX	XXX			0 XXX	•	XXX	XXX	7	XXX	
32 multiply lines (1) X (31)		MHz		XXX	XXX			0 XXX	-	XXX	XXX	10500	XXX	
33 delta f=abs((4)-(32))		MHz		XXX	XXX		1	500 XXX	-	XXX	XXX	500	XXX	
34 delta $f > (BT+BR)/2$		dB	line (33), fig 2.10	XXX	XXX		-	100 XXX		XXX	XXX	-100	XXX	
35 calculate minimum delta f		MHz	form A	XXX	XXX		XXX		0	XXX	XXX	XXX	0	
36 delta f>(BT+BR)/2		dB	line (35), fig 2.10	XXX	XXX		XXX		0	XXX	XXX	XXX	0	
EMI Frequency Corrected Summa	ary													
37 Add line 21 to line				25		30		34	36	25	30	34	36	
38 Total		dB		0		110		112	129	0	92	130	112	

Table A.3.8.b. AMTI and GMTI Frequency Correction

IF EMI Margin < -10dB, EMI Highly Improbable

				GMTI-RX IFF-TX		IFF-RX GMTI-TX				AMTI-TX RJ-HB-RX						
Line	Parameter	Symbol	Unit	Recommended Value/Equation If Needed	FIM	TIM	RIM	SIM	FIM	TIM	RIM	SIM	FIM	TIM	RIM	SIM
	A	MPLITUI	DE CULLI	NG	NA			DIN	NA		101/1	01111	NA		10111	51111
8	TX Power	PT (f OT)	dBm			200	XXX	XXX		200	XXX	XXX		200	XXX	XXX
9	TX Spurios Power Output	PT (f ST)	dBm	PT(f OT)-60dB	XXX	XXX	140	140	XXX	XXX	140	140	XXX	XXX	140	140
10	TX Antenna Gain in RX Direction	GTR (f)	dB	or 0dB		0	0	0		0	0	0		0	0	0
11	RX Antenna Gain in TX Direction	GRT (f)	dB	or 0dB		0	0	0		0	0	0		0	0	0
12	Propagation Loss Using Freq No.	L			#1	#1	#4	#2	#1	#1	#4	#2	#1	#1	#4	#2
			MHz			1030	10000	103		10000	1090	1000		1500	17	150
	Loss from Fig 1 (pg2.36,Vol7 (Function of freq.	No. 2.7, ) Duff , distance)	dB	all negative		61	82	41		82	61	58		57	20	40
13	Unintentional Power Available	PA (f)	dBm	PT + L	0	261	222	181	0	282	201	198	0	257	160	180
14	RX Fundamental Susceptibility	PR (f OR)	dBm	sensitivity		XXX	-100	XXX		XXX	-84	XXX		XXX	-50	XXX
15	RX Spurious Suscept	PR (f SR)	dBm	PR (f OR) + 80dB	XXX	-20	XXX	-20	XXX	-4	xxx	-4	xxx	30	xxx	30
16	Preliminary EMI Prediction		dB	line 13-14 or 13-15	0	281	322	201	0	286	285	202	0	227	210	150

#### Table A.3.9. GMTI-IFF, IFF-GMTI, and AMTI-HB Amplitude Culling

IF EMI margin < -10 dB, EMI Highly Improbable = STOP IF EMI margin > -10 dB, Start Frequency Culling

					GMTI-RX	TX	-	IFF-		IFF- GMT	-RX I-TX			AMT RJ-H	TI-TX B-RX	
Line	Parameter	Symbol	Unit	Recommended Value/Equation If Needed	FIM	TIM	RIM	SIM	FIM	TIM	RIM	SIM	FIM	TIM	RIM	SIM
FREQUENCY CULLING																
	Bandwidth Co	rrection														
17	TX PRF (if pulse)		pps			100	100	100		100	100	100		100	100	100
18	TX Bandwidth	BT		2/PI * t if pulse; t=width		1.91	1.91	1.9		6.366	6.37	6.37		6.37	6.4	6.4
19	RX Bandwidth	BR				10	10	10		8	8	8		25	25	25
20	Adjustment	lines 17 to 19	dB	Use Fig 2.8/2.9		-92	-92	-92		-92	-92	-92		-92	-92	-92
21	Bandwidth Corrected	EMI Margin	dB	line 16+20	0	189	230	109	0	194	193	110	0	135	118	58
	(BT+BR)/2				####	5.955	5.955	6	###	7.183	7.18	7.18	##	15.7	16	16
	IF EMI MARO	GIN <= -	10 dB, EMI HIGHL	Y IMPROBABLE = STOP												

## Table A.3.10.a. GMTI-IFF, IFF-GMTI, and AMTI-HB Frequency Culling

				,	ĺ	GMTI IFF-	-RX TX		•	IFF- GMT	RX I-TX			AMT RJ-H	'I-TX B-RX	
	Frequency Correction															
22	RX Local Oscillator Frequency	f LO	dBm			10100				1160				22		
23	RX Intermediate Frequency	f IF	dBm			100				70				5		
24	TX-RX Freq Separation:			delta f=abs((1)-(4)		XXX	XXX	XXX		XXX	XXX	XXX		XXX	XXX	XXX
25	delta f > (BT+BR)/2			line (24) & fig 2.10		ххх	ххх	ххх		ххх	XXX	ххх		ххх	XXX	ххх
26	f OT/ f LO +/- f IF to nearest integer				ххх	0	ххх	xxx	XXX	9	xxx	ххх	ххх	68	ххх	xxx
27	multiply lines (22) & (26)		MHz		XXX	0	XXX	XXX	XXX	10440	XXX	XXX	XXX	1496	XXX	XXX
28	delta f =abs((1)-(23)-(27))				XXX	1122	XXX	XXX	XXX	348	XXX	XXX	XXX	96	XXX	XXX
28	delta f =abs((1)+(23)-(27))				XXX	938	XXX	XXX	XXX	532	XXX	XXX	XXX	88	XXX	XXX
29	select smaller delta f from (28)		MHz		xxx	938	ххх	ххх	ххх	348	ххх	ххх	xxx	88	xxx	ххх
30	delta f>(BT+BR)/2		dB	line (29), fig 2.10	ххх	-100	ххх	xxx	ххх	-100	ххх	ххх	ххх	-100	ххх	xxx
31	calculate f OR/f OT to nearest integer				xxx	xxx	10	xxx	xxx	xxx	0	xxx	xxx	xxx	0	xxx
32	multiply lines (1) X (31)		MHz		XXX	XXX	10300	XXX	ххх	XXX	0	XXX	XXX	XXX	0	XXX
33	delta f=abs((4)-(32))		MHz		XXX	XXX	300	XXX	XXX	XXX	####	XXX	XXX	XXX	17	XXX
34	delta f > (BT+BR)/2		dB	line (33), fig 2.10	xxx	xxx	-100	ххх	ххх	ххх	-100	ххх	xxx	ххх	###	xxx
35	calculate minimum delta f		MHz	form A	XXX	XXX	XXX	0	XXX	XXX	XXX	0	XXX	XXX	XXX	0
36	delta f>(BT+BR)/2		dB	line (35), fig 2.10	xxx	xxx	xxx	0	xxx	xxx	xxx	0	xxx	xxx	xxx	0
	EMI Frequency Corrected Summary															
37	Add line 21 to line				25	30	34	36	25	30	34	36	25	30	34	36
38	Total		dB		0	89	130	109	0	94	93	110	0	35	18	58
			IF EN EMI	/I Margin < -10dB, Highly Improbable												

Table A.3.10.b. GMTI-IFF, IFF-GMTI, and AMTI-HB Frequency Correction

#### Alternate EMI Evaluation

## Table A.3.11.a. AMTI – GMTI and GMTI – IFF Analog EMC Prediction Combinations Out of Band Interference; separations of >10% operating frequency

Line Parameter	Symbol	Unit	AMTI-RX GMTI-TX	GMTI-RX AMTI-TX	GMTI-RX IFF-TX	IFF-RX GMTI-TX
Transmitter Harmonic to Receiver	Fundament	al• fR >	5111-17		111-1X	OWITI-TX
Transmitter fraimonie to Receiver	unuamenta	ai, iix >	119			
1 RX Frequency	fR	MHz	1500	10000	10000	1090
2 TX Frequency	fT	MHz	10000	1500	1030	10000
3(1)/(2) and round off to nearest integer	Ν		0	7	10	0
4 TX Harmonic Frequency; (3)*(2)	NfT	MHz	0	10500	10300	0
5 Frequency Separation; abs((4)-(1))		MHz	1500	500	300	1090
6 Receiver Bandwidth			10	10	10	8
If (5)>(6), No Harmonic Interference;	If		Na	Na	Na	Na
(3)<(6) Continue			INO	INO	NO	INO
7 TX Power	PT	dBm	200	200	200	200
8 Harmonic Corection (from table 8.2)		dB	0	0		
		uD	0	0	0	0
9 Harmonic Power	РТ	dBm	200	200	0 200	0 200
9 Harmonic Power 10 Propagation Constant	РТ	dBm	200 32	200 32	0 200 32	0 200 32
9 Harmonic Power 10 Propagation Constant 11 20 log dTR	РТ	dBm km	200 32 -27.767868	0 200 32 -27.767868	0 200 32 -27.7679	0 200 32 -27.7679
9 Harmonic Power 10 Propagation Constant 11 20 log dTR 12 20 log fR	РТ	dBm km MHz	200 32 -27.767868 63.5218252	0 200 32 -27.767868 80	0 200 32 -27.7679 80	0 200 32 -27.7679 60.74853
<ul> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> </ul>	PT L	dBm km MHz dB	200 32 -27.767868 63.5218252 67.7539568	0 200 32 -27.767868 80 84.2321317	0 200 32 -27.7679 80 84.23213	0 200 32 -27.7679 60.74853 64.98066
<ul> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> </ul>	PT L GR	dBm km MHz dB dB	200 32 -27.767868 63.5218252 67.7539568 60	0 200 32 -27.767868 80 84.2321317 60	0 200 32 -27.7679 80 84.23213 25	0 200 32 -27.7679 60.74853 64.98066 60
<ul> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> <li>15 Power Available at RX; (9)-(13)+(14)</li> </ul>	PT L GR	dBm km MHz dB dB dBm	200 32 -27.767868 63.5218252 67.7539568 60 192.246043	0 200 32 -27.767868 80 84.2321317 60 175.767868	0 200 32 -27.7679 80 84.23213 25 140.7679	0 200 32 -27.7679 60.74853 64.98066 60 195.0193
<ul> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> <li>15 Power Available at RX; (9)-(13)+(14)</li> <li>16 RX Susceptibility Level</li> </ul>	PT L GR PR	dBm km MHz dB dB dBm dBm	200 32 -27.767868 63.5218252 67.7539568 60 192.246043 -100	0 200 32 -27.767868 80 84.2321317 60 175.767868 -100	0 200 32 -27.7679 80 84.23213 25 140.7679 -100	0 200 32 -27.7679 60.74853 64.98066 60 195.0193 -84

Transmitter Fundamental to Receive	er Spurio	us; fT >	fR			
18(2)/(1) and round off to nearest integer	Р		6.66666667	0.15	0.103	9.174312
19 Local Oscillator Frequency	fLO	MHz	10100	10100	1160	22
20 Intermediate Frequency	fIF	MHz	60	100	100	70
21 abs(PfLO+/-fIF-fT; (18)*(19) + (20)-(2)			57393.3333	115	810.52	9728.165
abs(PfLO+/-fIF-fT; (18)*(19) - (20)-(2)			57273.3333	85	1010.52	9868.165
If $(21+)$ or $(21-) > (6)$ No Spurious Interface; If						
(21+) or (21-) <(6) Continue	line (6)		No	No	No	No
22 TX Power	PT	dBm	200	200	200	200
23 TX Antenna Gain	GT	DB	60	60	25	60
24 Propagation Constant			32	32	33	33
25 20 log dTR		km	-27.767868	-27.767868	-27.7679	-27.7679
26 20 log fT		MHz	80	63.5218252	60.25674	80
27 Propagation Loss; (24)+(25)+(26)	L	dB	84.2321317	67.7539568	65.48888	85.23213
28 Power Available at RX; (22)+(23)-(27)		dBm	175.767868	192.246043	159.5111	174.7679
29 RX Fundamental Susceptibitlity	PR	dBm	-100	-100	-100	-84
30 Spurious Correction (from Table 8.3)		dBm				
31 Spurious Susceptibility; (29)+(30)		dBm	-100	-100	-100	-84
32 Interference Margin; (28)-(31)	IM	dB	275.767868	292.246043	259.5111	258.7679

#### Table A.3.11.b. AMTI – GMTI and GMTI – IFF Analog EMC Prediction Combinations

IM<-10dB, EMI Highly Improbable -10dB<IM<10dB, EMI Marginal IM>10dB, EMI Probable

Line Parameter	Symbol	Unit	AMTI-TX HB-RX	AMTI-TX LB-RX	AMTI-TX SHF-RX	GMTI-TX HB-RX	GMTI-TX LB-RX	GMTI-TX SHF-RX
Transmitter Harmonic to Receiver Fund	amental; fR	l > f Tq						
1 RX Frequency	fR	MHz	17	0.08	1.8E+07	17	0.08	18000000
2 TX Frequency	fT	MHz	1500	1500	1500	10000	10000	10000
3(1)/(2) and round off to nearest integer	Ν		0	0	12000	0	0	1800
4 TX Harmonic Frequency; (3)*(2)	NfT	MHz	0	0	1.8E+07	0	0	18000000
5 Frequency Separation; abs((4)-(1))		MHz	17	0.08	0	17	0.08	0
6 Receiver Bandwidth			25	100	3E+07	25	100	3000000
If (5)>(6), No Harmonic Interference; If (5)<(6) Continue			Yes	Yes	Yes	Yes	Yes	Yes
7 TX Power	РТ	dBm	200	200	200	200	200	200
7 TX Power 8 Harmonic Corection (from table 8.2)	РТ	dBm dB	200 0	200 0	200 1	200 2	200 3	200 4
7 TX Power 8 Harmonic Corection (from table 8.2) 9 Harmonic Power	PT PT	dBm dB dBm	200 0 200	200 0 200	200 1 200	200 2 200	200 3 200	200 4 200
7 TX Power 8 Harmonic Corection (from table 8.2) 9 Harmonic Power 10 Propagation Constant	PT PT	dBm dB dBm	200 0 200 32	200 0 200 32	200 1 200 32	200 2 200 32	200 3 200 32	200 4 200 32
7 TX Power 8 Harmonic Corection (from table 8.2) 9 Harmonic Power 10 Propagation Constant 11 20 log dTR	PT PT	dBm dB dBm km	200 0 200 32 -37.646	200 0 200 32 -37.6461	200 1 200 32 -37.6461	200 2 200 32 -30.458	200 3 200 32 -30.458	200 4 200 32 -30.4576
7 TX Power 8 Harmonic Corection (from table 8.2) 9 Harmonic Power 10 Propagation Constant 11 20 log dTR 12 20 log fR	PT PT	dBm dB dBm km MHz	200 0 200 32 -37.646 24.609	200 0 200 32 -37.6461 -21.9382	200 1 200 32 -37.6461 145.105	200 2 200 32 -30.458 24.609	200 3 200 32 -30.458 -21.938	200 4 200 32 -30.4576 145.1055
<ul> <li>7 TX Power</li> <li>8 Harmonic Corection (from table 8.2)</li> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> </ul>	PT PT L	dBm dB dBm km MHz dB	200 0 200 32 -37.646 24.609 18.9629	200 0 200 32 -37.6461 -21.9382 -27.5843	200 1 200 32 -37.6461 145.105 139.459	200 2 200 32 -30.458 24.609 26.1514	200 3 200 32 -30.458 -21.938 -20.396	200 4 200 32 -30.4576 145.1055 146.6479
<ul> <li>7 TX Power</li> <li>8 Harmonic Corection (from table 8.2)</li> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> </ul>	PT PT L GR	dBm dB dBm km MHz dB dB	200 0 200 32 -37.646 24.609 18.9629 60	200 0 200 32 -37.6461 -21.9382 -27.5843 60	200 1 200 32 -37.6461 145.105 139.459 60	200 2 200 32 -30.458 24.609 26.1514 60	200 3 200 32 -30.458 -21.938 -20.396 60	200 4 200 32 -30.4576 145.1055 146.6479 60
<ul> <li>7 TX Power</li> <li>8 Harmonic Corection (from table 8.2)</li> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> <li>15 Power Available at RX; (9)-(13)+(14)</li> </ul>	PT PT L GR	dBm dB dBm km MHz dB dB dBm	200 0 200 32 -37.646 24.609 18.9629 60 241.037	200 0 200 32 -37.6461 -21.9382 -27.5843 60 287.584	200 1 200 32 -37.6461 145.105 139.459 60 120.541	200 2 200 32 -30.458 24.609 26.1514 60 233.849	200 3 200 32 -30.458 -21.938 -20.396 60 280.396	200 4 200 32 -30.4576 145.1055 146.6479 60 113.3521
<ul> <li>7 TX Power</li> <li>8 Harmonic Corection (from table 8.2)</li> <li>9 Harmonic Power</li> <li>10 Propagation Constant</li> <li>11 20 log dTR</li> <li>12 20 log fR</li> <li>13 Propagation Loss; (10)+(11)+(12)</li> <li>14 RX Antenna Gain</li> <li>15 Power Available at RX; (9)-(13)+(14)</li> <li>16 RX Susceptibility Level</li> </ul>	PT PT L GR PR	dBm dB dBm km MHz dB dB dBm dBm	200 0 200 32 -37.646 24.609 18.9629 60 241.037 -50	200 0 200 32 -37.6461 -21.9382 -27.5843 60 287.584 -50	200 1 200 32 -37.6461 145.105 139.459 60 120.541 -50	200 2 200 32 -30.458 24.609 26.1514 60 233.849 -50	200 3 200 32 -30.458 -21.938 -20.396 60 280.396 -50	200 4 200 32 -30.4576 145.1055 146.6479 60 113.3521 -50

## Table A.3.12.a. AMTI and GMTI Combinations with HB, LB, and SHF Analog EMC Prediction Out of Band Interference; separations of >10% operating frequency

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						0		
<b>Transmitter Fundamental to Receiver Sp</b>	urious; fT	' > fR						
18(2)/(1) and round off to nearest integer	Р		88.2353	18750	8.3E-05	588.235	125000	0.000556
19 Local Oscillator Frequency	fLO	MHz	0.11	3.8E+07	0	0	0	0
20 Intermediate Frequency	fIF	MHz	5	0.03	2E+07	5	0.03	2000000
21 abs(PfLO+/-fIF-fT; (18)*(19) + (20)-(2)			1485.29	7.1E+11	2E+07	9995	9999.97	19990000
abs(PfLO+/-fIF-fT; (18)*(19) - (20)-(2)			1495.29	7.1E+11	2E+07	10005	10000	20010000
If $(21+)$ or $(21-) > (6)$ No Spurious Interface: If $(21+)$ or								
(21-) <(6) Continue	line (6)		No	No	Yes	No	No	Yes
22 TX Power	PT	dBm	200	200	200	200	200	200
23 TX Antenna Gain	GT	DB	60	60	60	60	60	60
24 Propagation Constant			33	33	33	33	33	33
25 20 log dTR		km	-37.646	-37.6461	-37.6461	-30.458	-30.458	-30.4576
26 20 log fT		MHz	63.5218	63.5218	63.5218	80	80	80
27 Propagation Loss; (24)+(25)+(26)	L	dB	58.8758	58.8758	58.8758	82.5424	82.5424	82.54243
28 Power Available at RX; (22)+(23)-(27)		dBm	201.124	201.124	201.124	177.458	177.458	177.4576
29 RX Fundamental Susceptibitlity	PR	dBm	-50	-50	-50	-50	-50	-50
30 Spurious Correction (from Table 8.3)		dBm						
31 Spurious Susceptibility; (29)+(30)		dBm	-50	-50	-50	-50	-50	-50
32 Interference Margin; (28)-(31)	IM	dB	251.124	251.124	251.124	227.458	227.458	227.4576

## Table A.3.12.b. AMTI and GMTI Combinations with HB, LB, and SHF Analog EMC Prediction cansmitter Fundamental to Receiver Spurious: fT > fR

IM<-10dB, EMI Highly Improbable -10dB<IM<10dB, EMI Marginal IM>10dB, EMI

Probable

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13. SUPPLEMENTARY NOTES									
14. ABSTRACT									
The multi-mission aircraft (1	MMA) technical feas	ibility study	will look at re	placement of the aging fleet of C-					
135 and C-130 theater base	d command & cont	rol (C2) and	intelligence,	surveillance and reconnaissance					
(ISR) fleet. It is proposed	that the MMA be c	out-fitted to a	combine some	e or all the functions of existing					
AWACS, JSTARS, RIVET	JOINT, COMPASS	CALL, and A	ABCCC platfo	orms. It would also have links to					
other manned or unmanned	SR aircraft, as well	as satellites.	The objective	e of the proposed feasibility study					
is to examine the technical	risks involved in co	mbining mul	tiple function	s onto one aircraft that currently					
reside on separate aircraft.	This Thesis will spe	cifically foci	us on the risk	s that are due to electromagnetic					
interference between transm	itters and interference	e between a	ctive and pass	sive sensors. These risks will be					
outlined in detail and a reco	mmendation as to v	which function	ns could be c	combined with minimal technical					
risk will be made									
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